

## 电路理论（2）B考试范围

- ✓ 动态电路的输入—输出方程
- ✓ 线性一阶电路（三要素法）
- ✓ 单位阶跃响应、冲激响应
- ✓ 正弦稳态电路的分析（相量法、功率计算）
- ✓ 谐振
- ✓ 理想变压器、耦合电感
- ✓ 对称三相电路
- ✓ 谐波分析法

# 第11章 非正弦周期信号线性电路的稳态分析

## 一、非正弦周期电压电流的有效值的计算

$$u(t) = U_0 + \sum_{k=1}^{\infty} U_{km} \sin(k\omega t + \theta_k)$$

$$U = \sqrt{\frac{1}{T} \int_0^T u^2(t) dt}$$

$$U = \sqrt{U_0^2 + \sum_{k=1}^{\infty} U_k^2}$$

谐波电压有效值

$$i(t) = I_0 + \sum_{k=1}^{\infty} I_{km} \sin(k\omega t + \theta_k)$$

$$I = \sqrt{\frac{1}{T} \int_0^T i^2(t) dt}$$

$$I = \sqrt{I_0^2 + \sum_{k=1}^{\infty} I_k^2}$$

谐波电流有效值

## 二、非正弦周期电路有功功率的计算

$$i(t) = I_0 + \sum_{k=1}^{\infty} I_{km} \sin(k\omega t + \theta_{ik})$$

$$u(t) = U_0 + \sum_{k=1}^{\infty} U_{km} \sin(k\omega t + \theta_{uk})$$

$$P = \frac{1}{T} \int_0^T u(t)i(t)dt$$

直流分量和同频率的  
电压电流产生的功率

$$P = U_0 I_0 + \sum_{k=1}^{\infty} I_K U_K \cos(\theta_{uk} - \theta_{ik})$$

例：

$$u(t) = 20 + 5\sqrt{10} \sin(\omega t + 60^\circ) + 10\sqrt{2} \sin(3\omega t + 75^\circ) V$$

$$i(t) = 2 + \sin(3\omega t + 30^\circ)$$

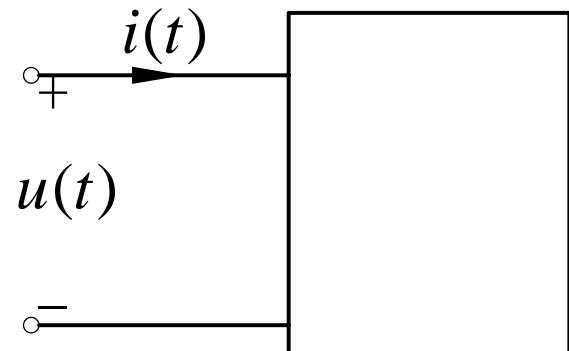
$$U = \sqrt{U_0^2 + \sum_{k=1}^{\infty} U_k^2} = \sqrt{20^2 + (5\sqrt{10}/\sqrt{2})^2 + (10 \times \sqrt{2}/\sqrt{2})^2}$$

$$= \sqrt{400 + 125 + 100} = 25V$$

$$I = \sqrt{2^2 + (1/\sqrt{2})^2} = \sqrt{4.5} = 2.12A$$

$$P = 20 \times 2 + 10 \times 1/\sqrt{2} \times \cos(75^\circ - 30^\circ)$$

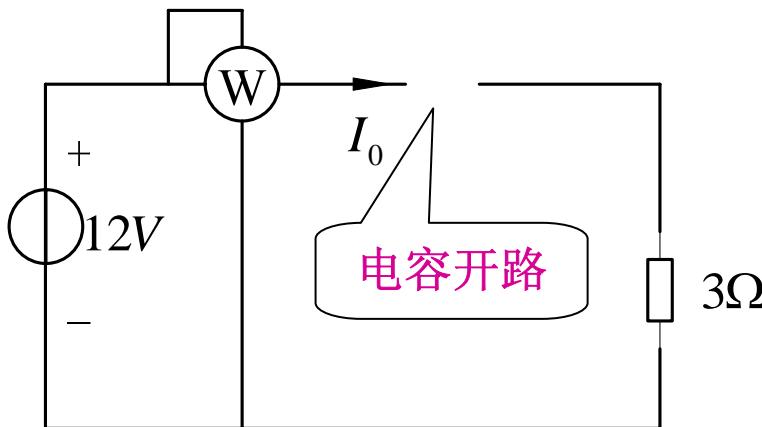
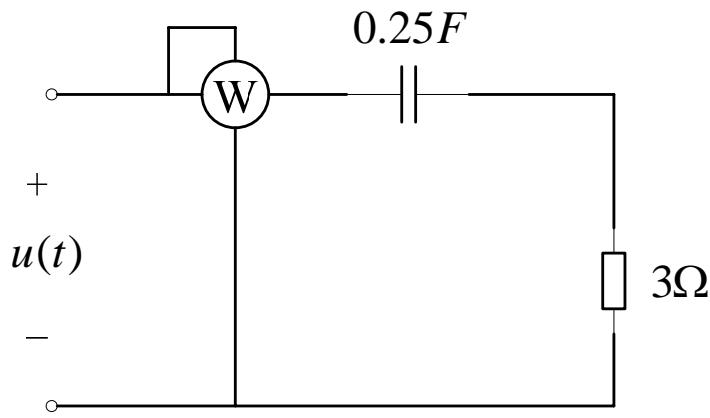
$$= 40 + 5 = 45W$$



同频率电压  
电流

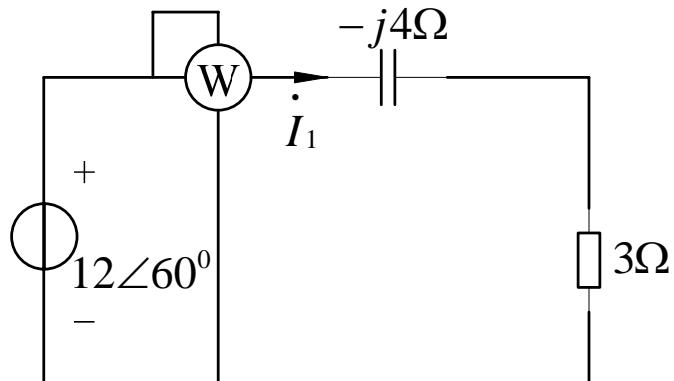
例:  $u(t) = 12 + 12\sqrt{2} \sin(t + 60^\circ) V$

解 1) 12V直流电源作用



$$I_0 = 0A$$

2) 交流电源作用 (相量模型)



$$i_1(t) = 2.4\sqrt{2} \sin(t + 113^\circ)$$

3) 时域叠加

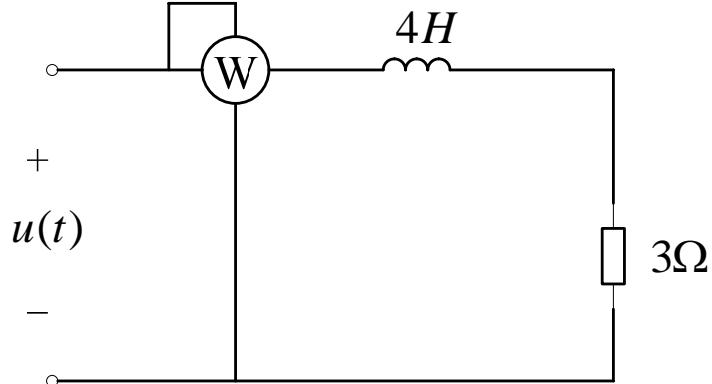
$$i(t) = I_0 + i_1(t) = 2.4\sqrt{2} \sin(t + 113^\circ)$$

4) 功率

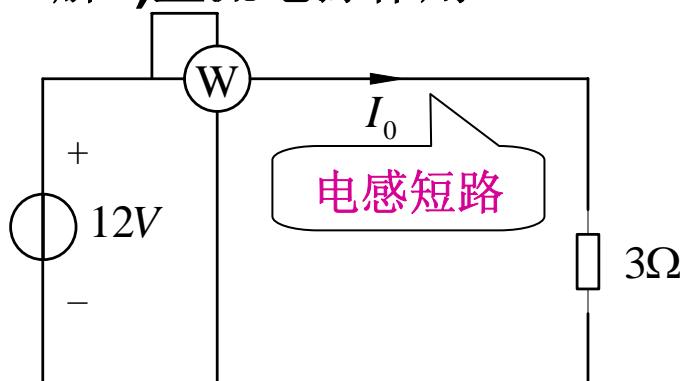
$$P = P_0 + P_1 = 12 \times 0 + 12 \times 2.4 \times \cos(-53^\circ) = 17$$

$$\dot{I}_1 = \frac{12\angle 60^\circ}{3 - 4j} = \frac{12\angle 60^\circ}{5\angle -53^\circ} = 2.4\angle 113^\circ$$

例:  $u(t) = 12 + 15\sqrt{2} \sin(t + 60^\circ) V$  2) 交流电源作用 (相量模型)

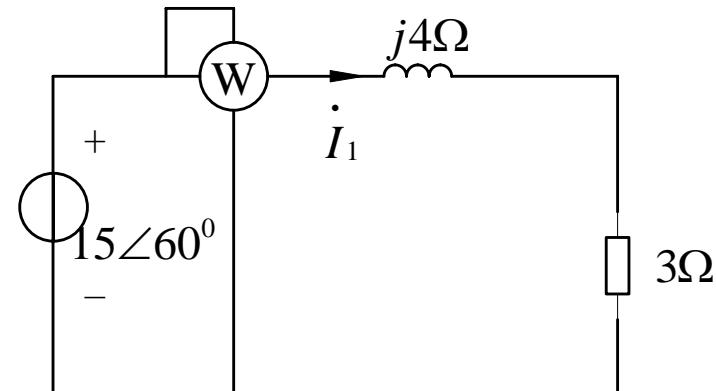


解1) 直流电源作用



$$I_0 = 12/3 = 4A$$

$$P_0 = 12 \times 4 = 48W$$



$$\dot{I} = \frac{15\angle 60^\circ}{3 + j4} = \frac{15\angle 60^\circ}{5\angle 53^\circ} = 3\angle 7^\circ A$$

$$i_1(t) = 3\sqrt{2} \sin(t + 7^\circ)$$

$$P_1 = 15 \times 3 \times \cos(53^\circ) = 27W$$

3) 响应叠加

$$i(t) = I_0 + i_1 = 4 + 3\sqrt{2} \sin(t + 7^\circ)$$

4) 功率

$$P = P_0 + P_1 = 48 + 27 = 75W$$

### 三、非正弦周期线性电路分析

叠加定理（时域形式的叠加）

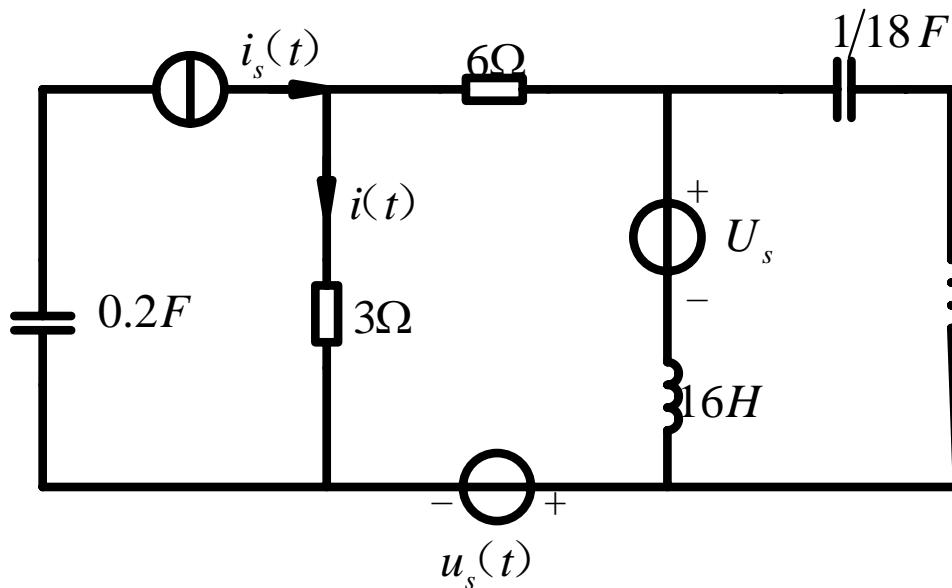
直流电源和正弦波电源作用分别求解

不同频率下感抗和阻抗的不同

注意  
谐振

$$X_C = -\frac{1}{k\omega C} \quad X_L = k\omega L$$

例1：



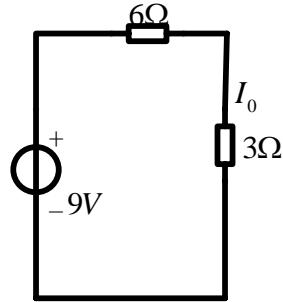
$$u_s(t) = 5 \sin(t + 90^\circ)$$

$$2H i_s(t) = 3\sqrt{2} \sin(3t + 30^\circ)$$

$$U_s = 9$$

解：

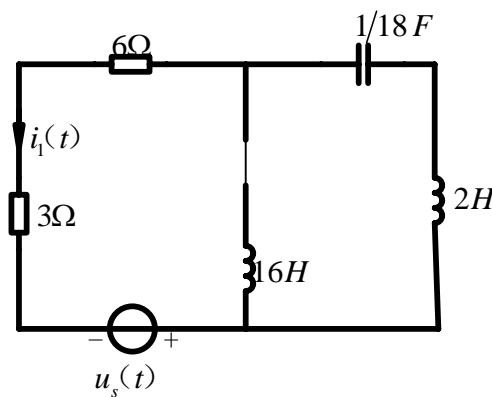
1) 直流电压源作用



$$I_0 = \frac{9}{3+6} = 1$$

$$P_0 = 3 \times 1^2 = 3W$$

2) 基波电源作用 (频率为1rad/s)

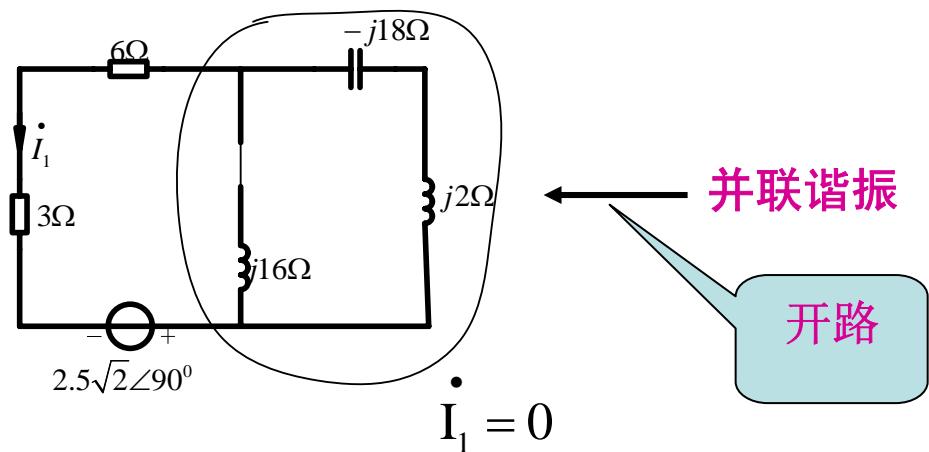


$$X_{L2} = 1 \times 16 = 16\Omega$$

$$X_{L1} = 1 \times 2 = 2\Omega$$

$$X_C = -\frac{1}{\omega C} = -18$$

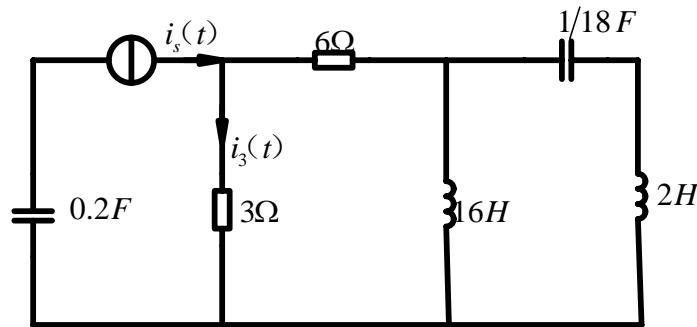
$$\dot{U}_s = 2.5\sqrt{2}\angle 90^\circ$$



$$\dot{I}_1 = 0$$

$$P_1 = 3 \times 0^2 = 0W$$

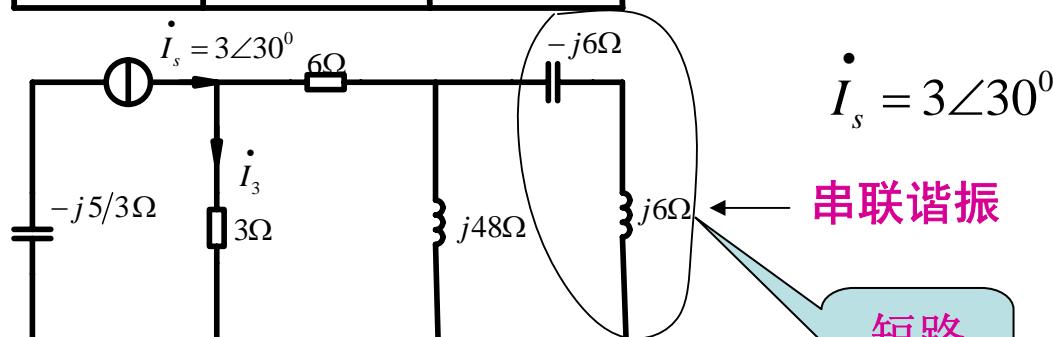
### 3) 3次谐波源作用



$$X_{C1} = -\frac{1}{3C} = -6\Omega \quad X_{C2} = -\frac{5}{3}\Omega$$

$$X_{L2} = 3 \times 16 = 48\Omega$$

$$X_{L1} = 3 \times 2 = 6\Omega$$



$$\dot{I}_s = 3\angle 30^\circ$$

串联谐振

短路

$$\dot{I}_3 = \frac{6}{(3+6)} \times 3\angle 30^\circ = 2\angle 30^\circ \quad i_3(t) = 2\sqrt{2} \sin(3t + 30^\circ) \quad P_3 = 3I_3^2 = 3 \times 2^2 = 12W$$

### 4) 将结果时域进行叠加

$$i(t) = I_0 + i_1(t) + i_3(t) = 1 + 2\sqrt{2} \sin(3t + 30^\circ) A$$

$$I = \sqrt{I_0^2 + I_1^2 + I_3^2} = \sqrt{1+4} = \sqrt{5}A$$

$$P = 3 + 12 = 15W$$

# 第10章 三相电路

## 一、三相对称电源

$$\dot{U}_{AN} = U_{ph} \angle 0^\circ \quad \dot{U}_{BN} = U_{ph} \angle -120^\circ \quad \dot{U}_{CN} = U_{ph} \angle 120^\circ$$

## 二、电源和负载的接法

星形接线和三角形接线方式及线相电压电流的定义

## 三、对称的概念

幅值相同，频率相同，相位差120度

## 四、线相电压线相电流关系（三相对称）

星型接法

$$\dot{U}_{AB} = \sqrt{3} \dot{U}_A \angle 30^\circ$$

$$U_l = \sqrt{3} U_{ph}$$

三角型接法

$$\dot{I}_{AB} = \frac{\dot{U}_{AB}}{Z_\Delta} \quad \dot{I}_A = \sqrt{3} \dot{I}_{AB} \angle -30^\circ$$

$$U_l = U_{ph} \quad I_l = \sqrt{3} I_{ph}$$

线电压相电压相等

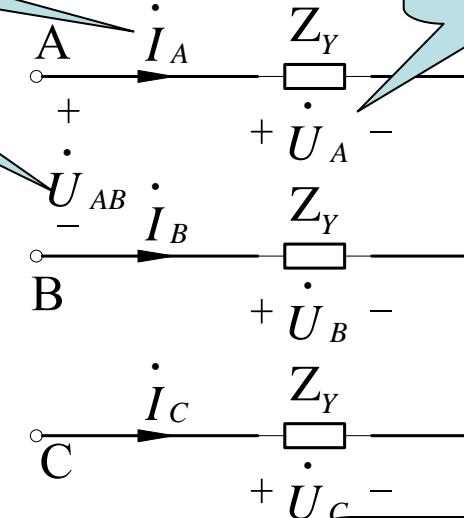
$$\dot{I}_A = \frac{\dot{U}_A}{Z_Y}$$

$$I_l = I_{ph}$$

线（相）电流

线电压

线电流相电  
流相等

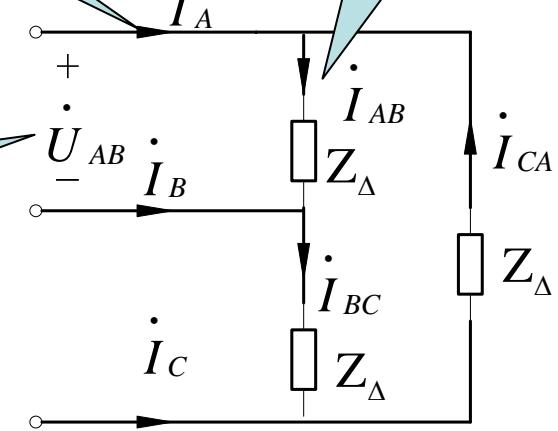


相电压

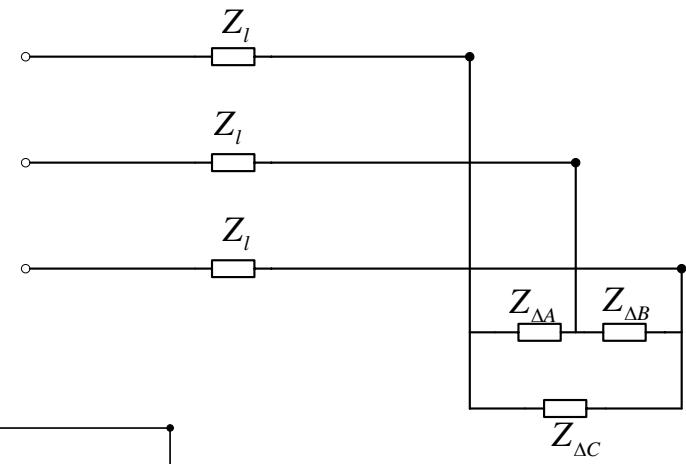
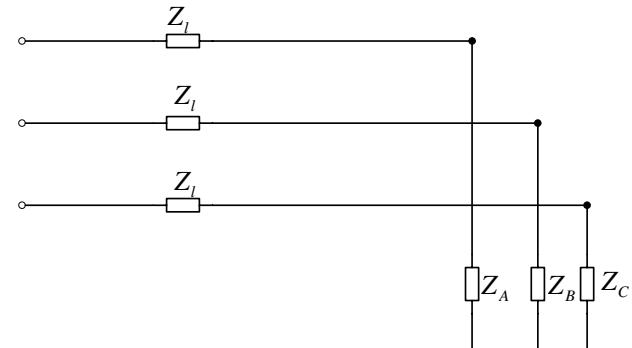
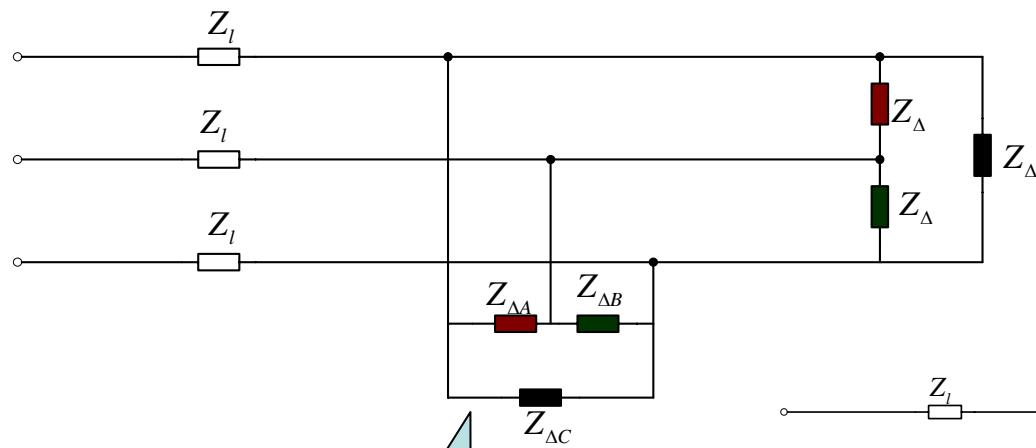
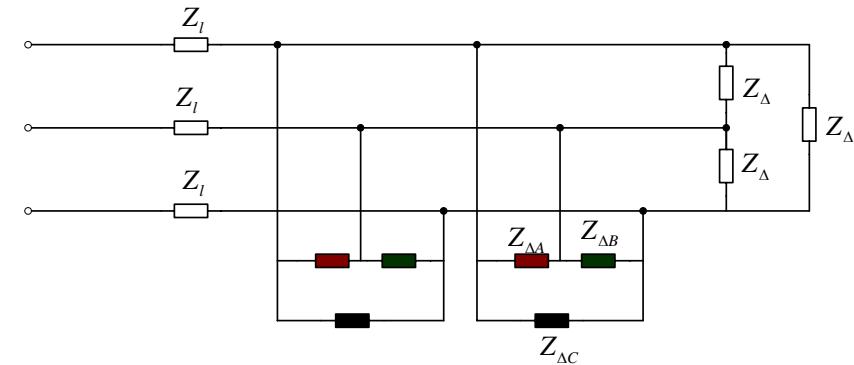
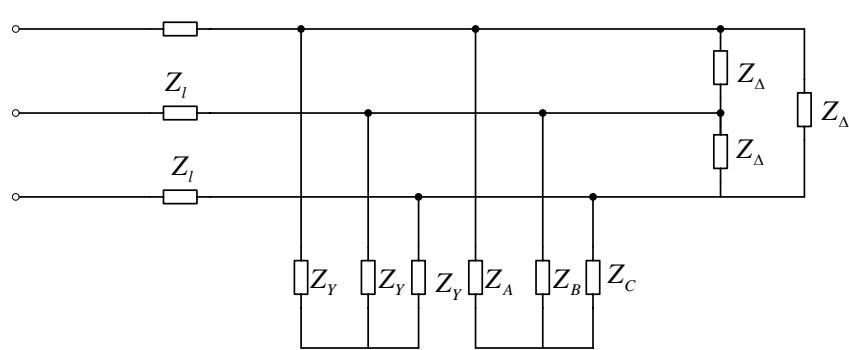
线电流

相电流

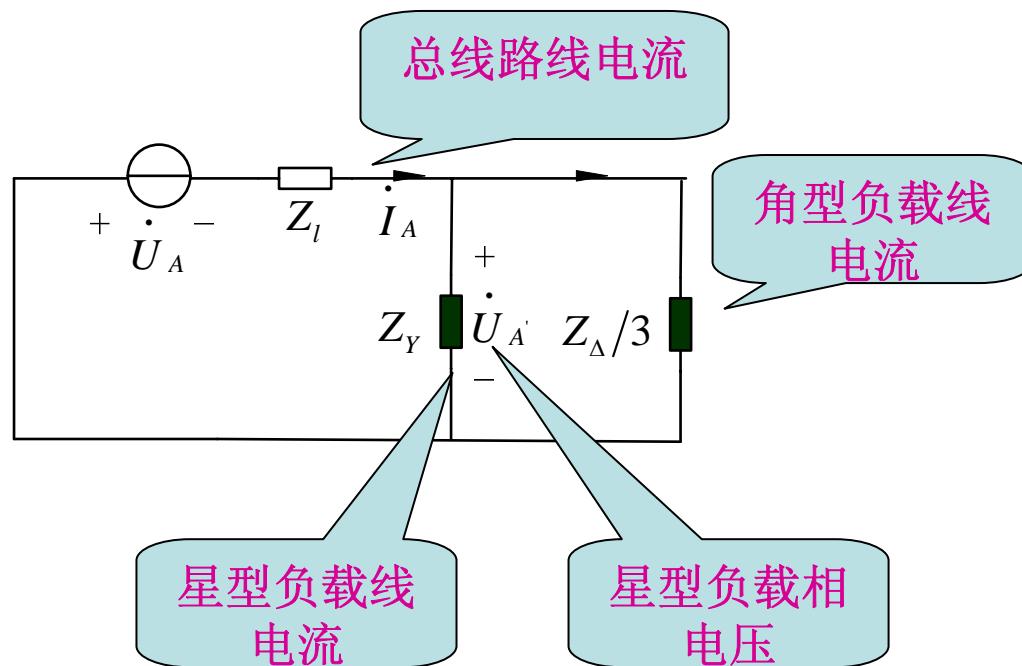
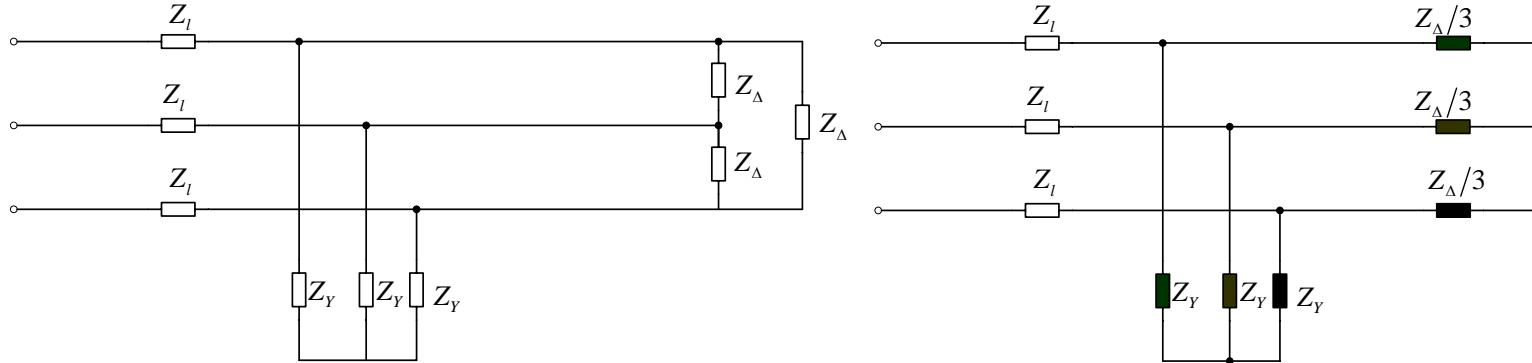
线（相）电压



# 五、多相不对称负载



# 六、三相对称电路等效



## 七、对称三相电路计算

将三相对称电路化为单相电路，根据对称性写出其它相的电压电流

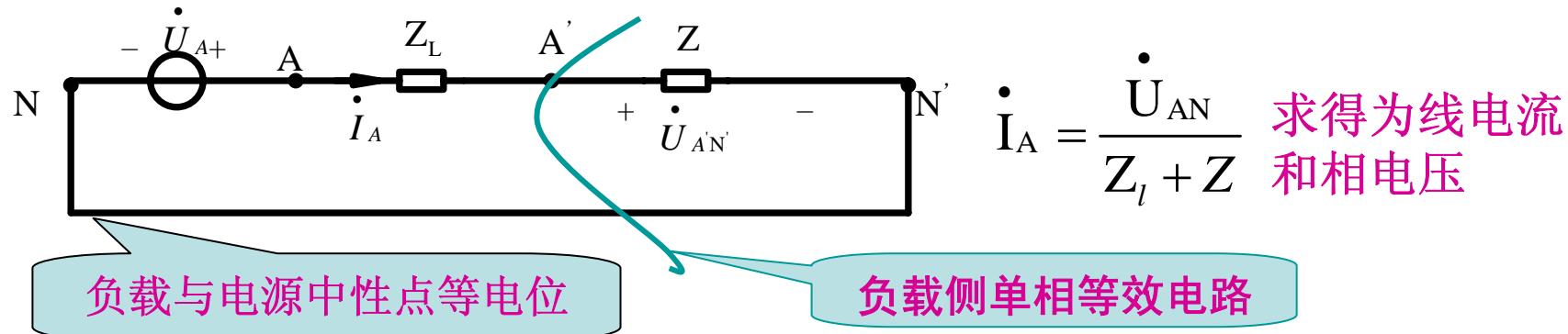
1) 根据线相电压关系得出星接电源或负载相电压

2) 选参考正弦量

3) 根据星三角变换将三角负载化为星型负载(所有的中性点为等电位点)

$$Z_Y = \frac{Z_\Delta}{3}$$

4) 根据单相电路计算出电流和电压，根据线相电压电流关系求解出待求量



# 八、对称三相电路功率计算

根据相电压和相电流计算

$$P = 3U_{ph} \times I_{ph} \times \cos \theta \quad Q = 3U_{ph} \times I_{ph} \times \sin \theta$$

$$I_{ph} = \frac{U_{ph}}{|Z|}$$

根据线电压线电流计算

$$P = \sqrt{3}U_l \times I_l \times \cos \theta \quad Q = \sqrt{3}U_l \times I_l \times \sin \theta$$

两表法测量与计算

线电压  
反向

$$P_1 = U_{AC} \times I_A \times \cos(\theta - 30^\circ)$$

线电压反向计算公式

$$Q = \sqrt{3}(P_1 - P_2)$$

线电压未反向

$$P_2 = U_{BC} \times I_B \times \cos(\theta + 30^\circ)$$

线电压未反向计算公式

$$P = P_1 + P_2$$

$$\theta = \Phi_Z$$

等效负载阻抗的辐角

待计算侧相电压和相电流的相位差

功率因数的计算

$$\theta = \theta_{u\_ph} - \theta_{i\_ph}$$

$$Z_{eq} = R + jX$$

$$\cos \theta = \frac{R}{\sqrt{R^2 + X^2}}$$

$$\sin \theta = \frac{X}{\sqrt{R^2 + X^2}}$$

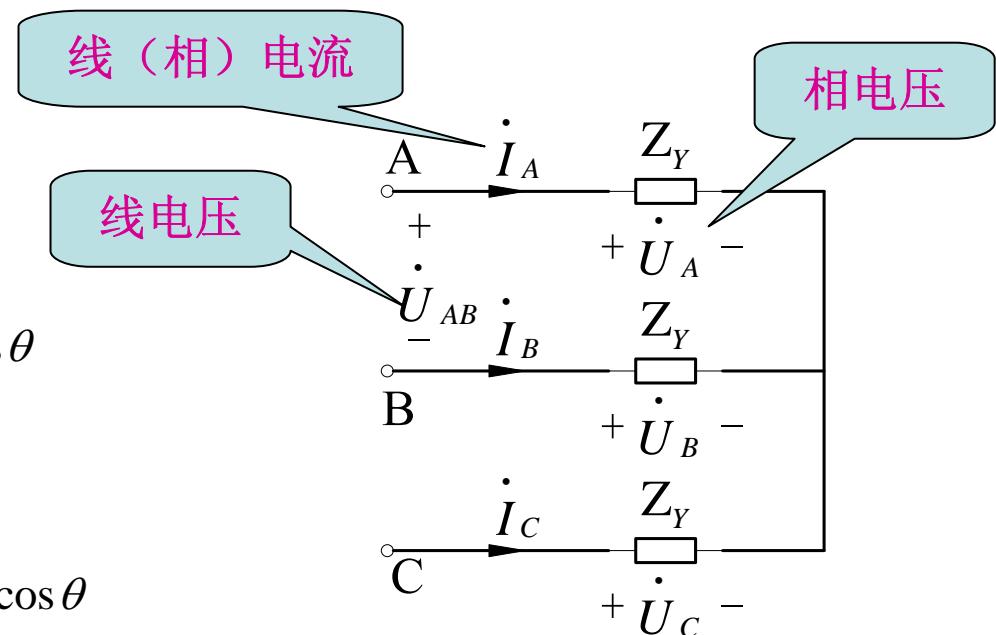
## 星型接法

按照相电压相电流计算

$$P = 3U_{Ph} \times I_{Ph} \times \cos \theta = 3U_A \times I_A \times \cos \theta$$

按照线电压线电流计算

$$P = \sqrt{3}U_L \times I_L \times \cos \theta = \sqrt{3}U_{AB} \times I_A \times \cos \theta$$



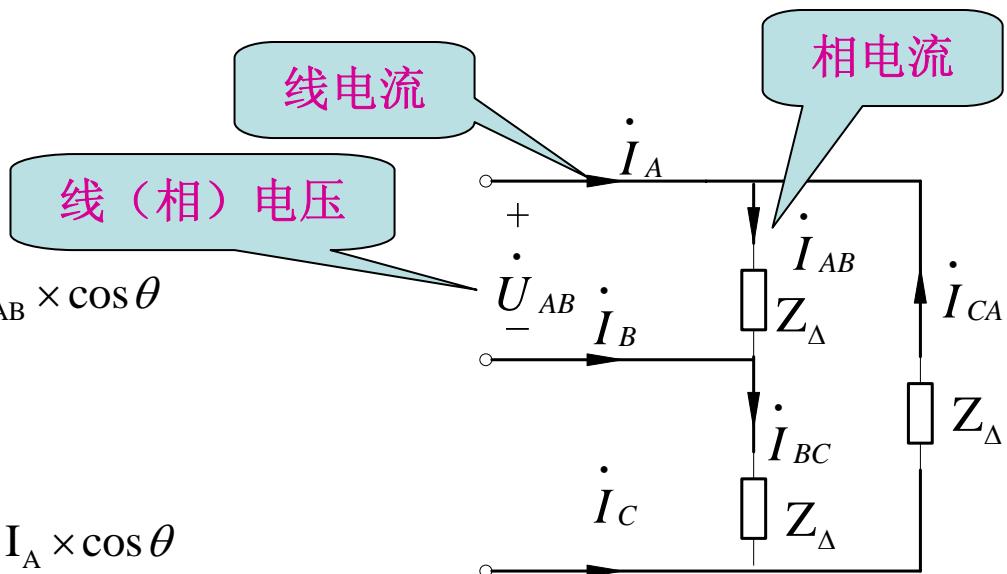
## 三角型接法

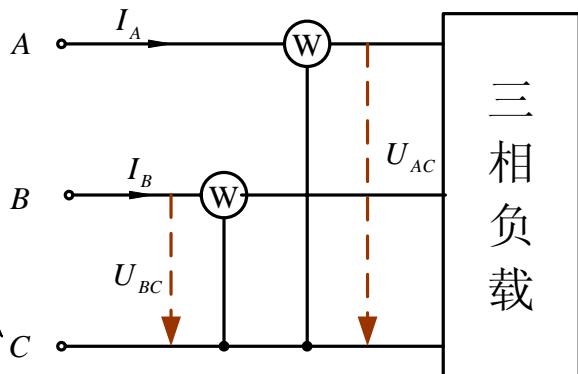
按照相电压相电流计算

$$P = 3U_{Ph} \times I_{Ph} \times \cos \theta = 3U_{AB} \times I_{AB} \times \cos \theta$$

按照线电压线电流计算

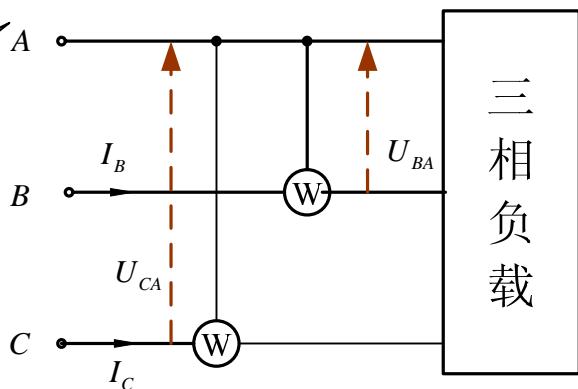
$$P = \sqrt{3}U_L \times I_L \times \cos \theta = \sqrt{3}U_{AB} \times I_A \times \cos \theta$$





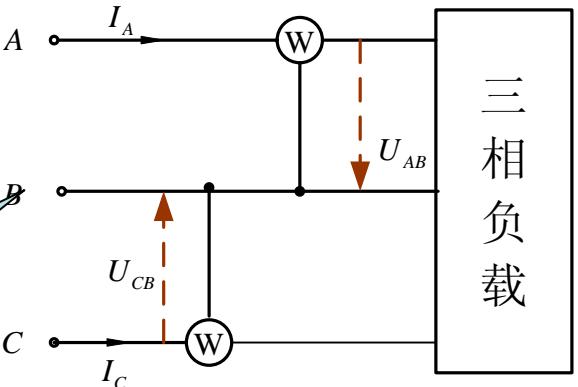
$$\begin{cases} P_1 = U_{AC} \times I_A \times \cos(\theta - 30^\circ) \\ P_2 = U_{BC} \times I_B \times \cos(\theta + 30^\circ) \end{cases}$$

线电压  
反向



$$\begin{cases} P_1 = U_{BA} \times I_B \times \cos(\theta - 30^\circ) \\ P_2 = U_{CA} \times I_C \times \cos(\theta + 30^\circ) \end{cases}$$

线电压  
未反向



$$\begin{cases} P_1 = U_{CB} \times I_C \times \cos(\theta - 30^\circ) \\ P_2 = U_{AB} \times I_A \times \cos(\theta + 30^\circ) \end{cases}$$

无功功  
率

$$Q = \sqrt{3} (P_1 - P_2)$$

例：三相对称系统参数如下： $U_{AB} = 380V$   $Z = 8 + j6\Omega$

解 1) 电源为星型接线则，令

$$\dot{U}_A = \frac{380}{\sqrt{3}} \angle 0^\circ = 220 \angle 0^\circ$$

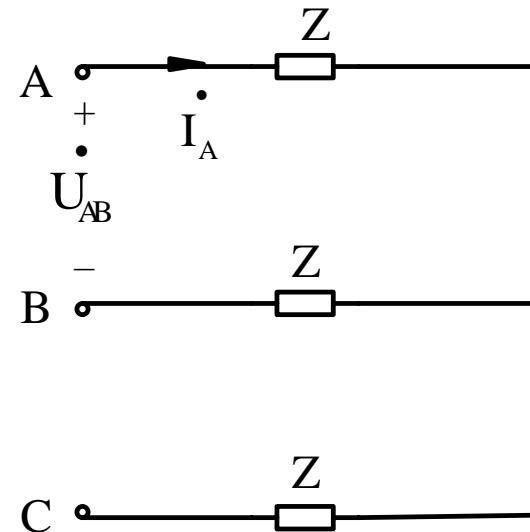
$$\dot{I}_A = \frac{\dot{U}_A}{Z} = \frac{220 \angle 0^\circ}{8 + 6j} = \frac{220 \angle 0^\circ}{10 \angle 37^\circ} = 22 \angle -37^\circ$$

$$\sin \theta = 6 / \sqrt{8^2 + 6^2} = 0.6 \quad \cos \theta = 8 / \sqrt{8^2 + 6^2} = 0.8$$

2) 或者  $\theta = 0^\circ - (-) = 37^\circ$

$$P = \sqrt{3} U_l \times I_l \times \cos \theta = \sqrt{3} \times 380 \times 22 \times 0.8 = 11616W$$

$$Q = \sqrt{3} U_l \times I_l \times \sin \theta = \sqrt{3} \times 380 \times 22 \times 0.6 = 8712 \text{ var}$$

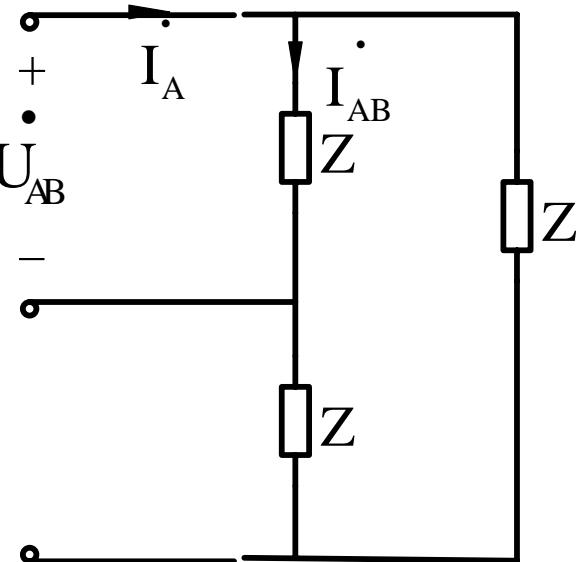


例：三相对称系统参数如下：  $U_{AB} = 380V$      $Z = 8 + j6\Omega$

解 1) 令  $\dot{U}_{AB} = 380\angle 0^\circ$

$$\dot{I}_{AB} = \frac{\dot{U}_{AB}}{Z} = \frac{380\angle 0^\circ}{8 + 6j} = \frac{380\angle 0^\circ}{10\angle 37^\circ} = 38\angle -37^\circ$$

$$\dot{I}_A = 38\sqrt{3}\angle -37^\circ - 30^\circ = 38\sqrt{3}\angle -67^\circ$$



2)

$$\sin \theta = 6 / \sqrt{8^2 + 6^2} = 0.6 \quad \cos \theta = 8 / \sqrt{8^2 + 6^2} = 0.8$$

$$P = \sqrt{3} U_l \times I_l \times \cos \theta = \sqrt{3} \times 380 \times 38\sqrt{3} \times 0.8 = 346560W$$

$$Q = \sqrt{3} U_l \times I_l \times \sin \theta = \sqrt{3} \times 380 \times 38\sqrt{3} \times 0.6 = 722000 \text{ var}$$

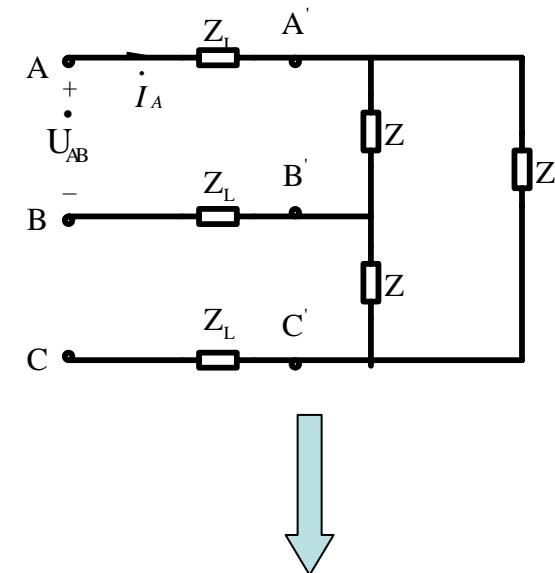
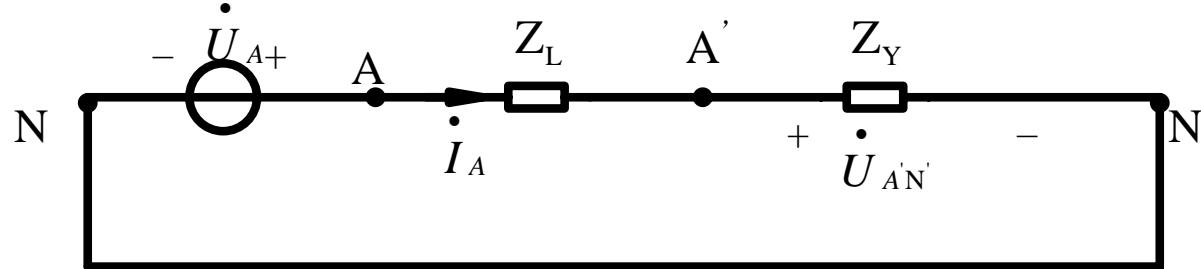
三相对称系统参数如下:  $U_{AB} = 380V$   $Z_L = 5 + j2\Omega$   $Z = 9 + j12\Omega$

求电路的有功和无功功率

解 1) 求参数画单相等值电路

$$Z_Y = \frac{9+12j}{3} = 3+j4\Omega$$

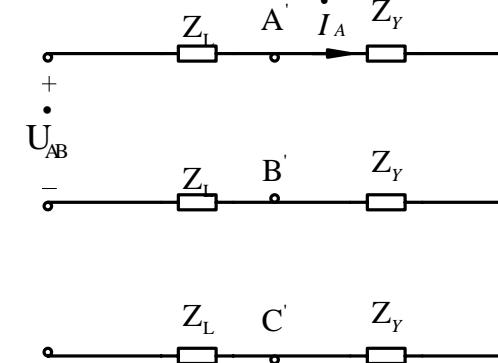
$$U_A = 380/\sqrt{3} = 220 \quad \text{令 } U_A = 220\angle 0^\circ$$

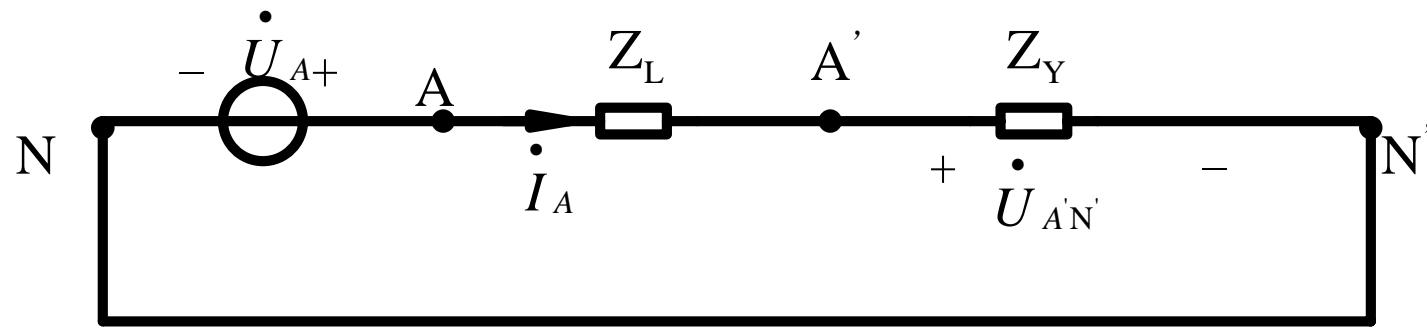


$$\dot{I}_A = \frac{\dot{U}_A}{Z_L + Z_Y} = \frac{220\angle 0^\circ}{3 + 4j + 5 + 2j}$$

$$= \frac{220\angle 0^\circ}{10\angle 37^\circ} = 22\angle -37^\circ$$

单相  
等值  
电路





$$2) \quad Z = Z_L + Z_Y = 3 + 4j + 5 + 2j = 8 + 6j \Omega$$

$$\sin \theta = 6 / \sqrt{8^2 + 6^2} = 0.6$$

$$\cos \theta = 8 / \sqrt{8^2 + 6^2} = 0.8$$

或  $\theta = 0 - (-37^\circ) = 37^\circ$

待计算电路相电压和  
相电流的相位差

3)

$$P = \sqrt{3} U_l \times I_l \times \cos \theta = \sqrt{3} \times 380 \times 22 \times 0.8 = 11616 \text{W}$$

$$Q = \sqrt{3} U_l \times I_l \times \sin \theta = \sqrt{3} \times 380 \times 22 \times 0.6 = 8712 \text{var}$$

已知负载端线电压为380V，线电流为2A负载功率因数为0.8（感性），线路阻抗为  $Z_l = 4 + j3\Omega$

求电源线电压  $\dot{U}_{AB}$   $\dot{U}_{BC}$   $\dot{U}_{CA}$

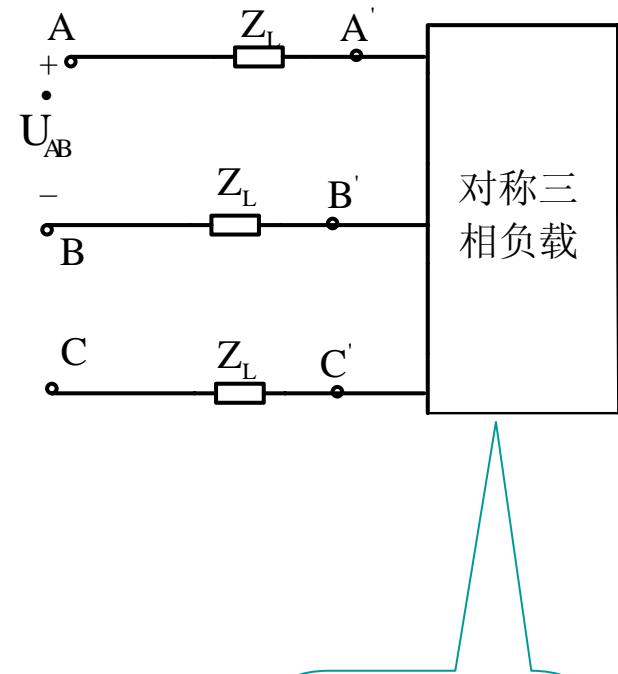
求电源提供的有功功率无功功率和视在功率

解：1)  $\dot{U}_{AB} = 380V$   $\dot{U}_{AN} = \frac{380}{\sqrt{3}} = 220V$

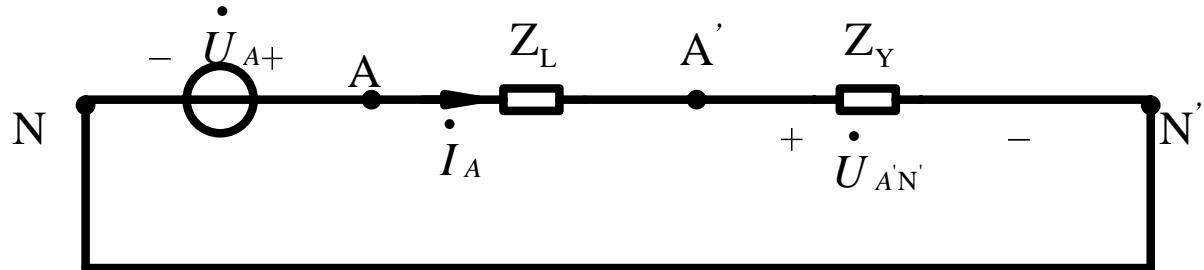
令  $\dot{U}_{AN} = 220\angle 0^\circ V$

功率因数为0.8（感性）  $\theta = 37^\circ$

则  $\dot{I}_A = \frac{\dot{U}_{AN}}{Z} = \frac{220\angle 0^\circ}{|Z|\angle 37^\circ} = 2\angle -37^\circ$



假设负载为  
星形接线



$$\dot{U}_{AN} = \dot{I}_A \times Z_l + \dot{U}_{A'N'} V$$

$$\dot{U}_{AN} = 2\angle -37^0 \times (4 + 3j) + 220\angle 0^0 = 230\angle 0^0 V$$

$$\dot{U}_{AB} = \sqrt{3} 230\angle 30^0 = 398.36\angle 30^0 V$$

3)

$$P = \sqrt{3} U_l \times I_l \times \cos \theta = \sqrt{3} \times 398.36 \times 2 \times 0.8 = 1104 W$$

$$Q = \sqrt{3} U_l \times I_l \times \sin \theta = \sqrt{3} \times 398.36 \times 2 \times 0.6 = 828 \text{ var}$$

$$S = \sqrt{3} U_l \times I_l = \sqrt{3} \times 398.36 \times 2 = 1380 \text{ VA}$$

已知电动机功率为2.5kW，功率因数为0.866，对称三相电源线电压为380V，求两个功率表的读数。

解：1)

$$P = \sqrt{3}U_l \times I_l \times \cos \theta$$

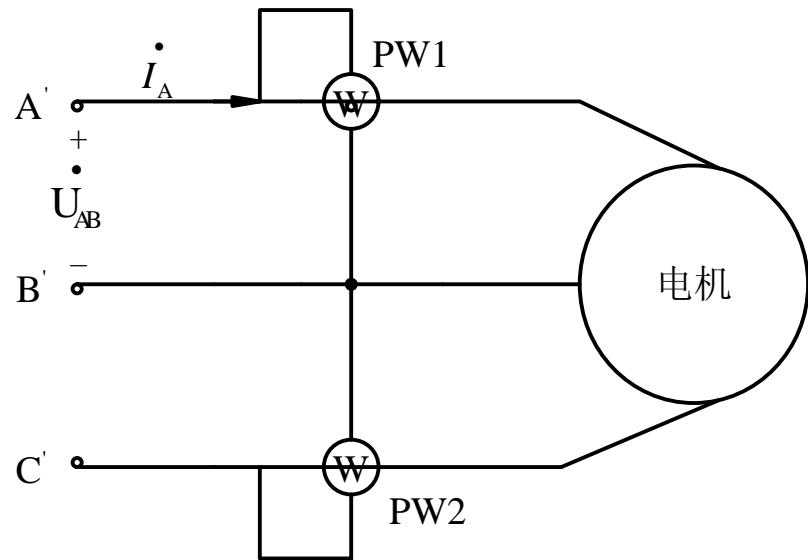
$$= \sqrt{3} \times 380 \times I_l \times 0.866 = 2500 \text{W}$$

$$I_l = \frac{2500}{\sqrt{3} \times 380 \times 0.866} = 4.386 \text{A}$$

2)  $\cos \theta = 0.866 \quad \theta = 30^\circ$

$$P_1 = U_{AB} \times I_A \times \cos(\theta + 30^\circ) = 380 \times 4.386 \times 0.5 = 833.34 \text{W}$$

$$P_2 = U_{CB} \times I_C \times \cos(\theta - 30^\circ) = 380 \times 4.386 \times 1 = 1666.68 \text{W}$$



# 没有线路阻抗的不对称电路

已知各个负载上的电流为8A，求Ic

解：以uab为参考相量

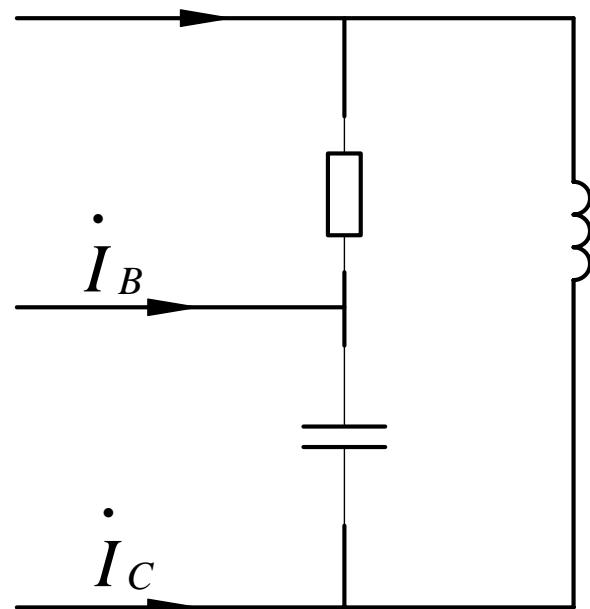
$$\dot{U}_{AB} = U_L \angle 0^\circ \quad \dot{I}_{AB} = \frac{U_L \angle 0^\circ}{R} = 8 \angle 0^\circ$$

$$\dot{U}_{BC} = U_L \angle -120^\circ$$

$$\dot{I}_{BC} = \frac{U_L \angle -120^\circ}{\frac{1}{\omega C} \angle -90^\circ} = 8 \angle -30^\circ$$

$$\dot{U}_{CA} = U_L \angle 120^\circ \quad \dot{I}_{CA} = \frac{U_L \angle 120^\circ}{\omega L \angle 90^\circ} = 8 \angle 30^\circ$$

$$\dot{I}_C = \dot{I}_{CA} - \dot{I}_{BC} = 8 \angle 30^\circ - 8 \angle -30^\circ = 8 \angle 90^\circ$$



已知:  $\dot{U}_{AB} = 173\angle 30^\circ$

$$Z_1 = 30 + 40j\Omega \quad Z_2 = 30\Omega$$

求功率表的示数

解: 1) 求参数画单相等值电路

$$Z_Y = \frac{30}{3} = 10\Omega \quad \dot{U}_A = \frac{173}{\sqrt{3}} = 100\angle 0^\circ$$

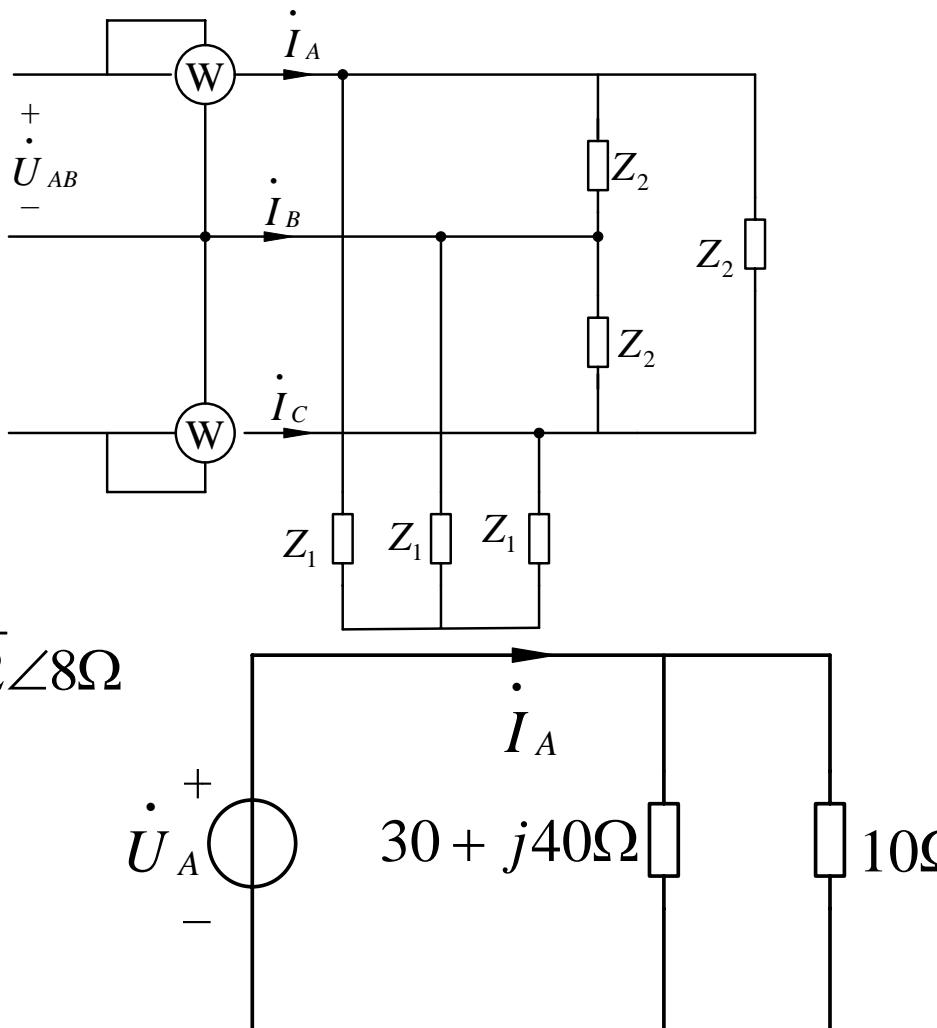
$$Z = \frac{(30 + 40j)10}{40 + j40} = \frac{50\angle 53^\circ}{4\sqrt{2}\angle 45^\circ} = 6.25\sqrt{2}\angle 8^\circ$$

$$\dot{I}_A = \frac{\dot{U}_A}{Z} = \frac{100\angle 0^\circ}{6.25\sqrt{2}\angle 8^\circ} = 8\sqrt{2}\angle -8^\circ$$

2)  $\theta = 8^\circ$

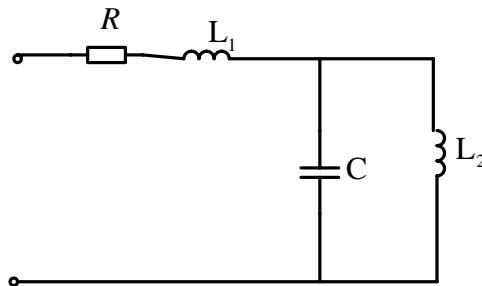
$$P_1 = U_{AB} \times I_A \times \cos(\theta + 30^\circ) = 173 \times 8\sqrt{2} \times 0.8 = 1565W$$

$$P_2 = U_{CB} \times I_C \times \cos(\theta - 30^\circ) = 173 \times 8\sqrt{2} \times 0.927 = 1814W$$



# 第9章 谐振与互感

## 一、谐振频率的计算



$$Z = R + j\omega L_1 + \frac{j\omega L_2 \times \frac{1}{j\omega C}}{j\omega L_2 - j\frac{1}{\omega C}}$$

等效阻抗或导纳的虚部为零

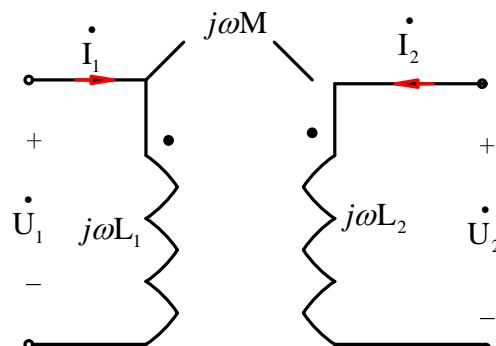
$$\omega_1 = \frac{1}{\sqrt{L_2 C}}$$

$$\omega_2 = \sqrt{\frac{L_2 + L_1}{L_2 L_1 C}}$$

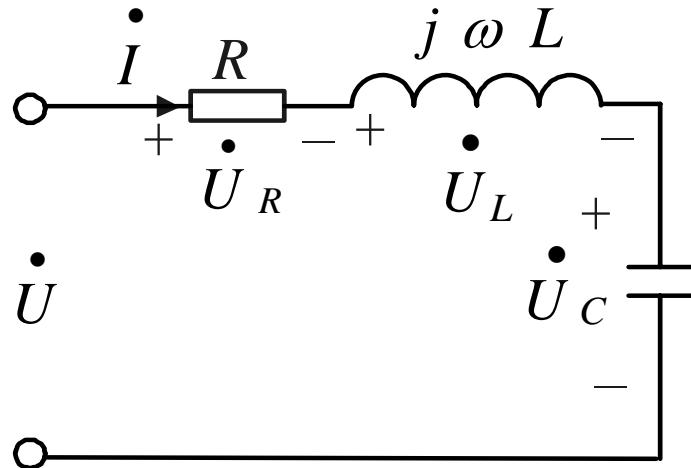
## 二、耦合电感电压电流关系

$$\dot{U}_1 = j\omega L_1 \times \dot{I}_1 + j\omega M \times \dot{I}_2$$

$$\dot{U}_2 = j\omega L_2 \times \dot{I}_2 + j\omega M \times \dot{I}_1$$

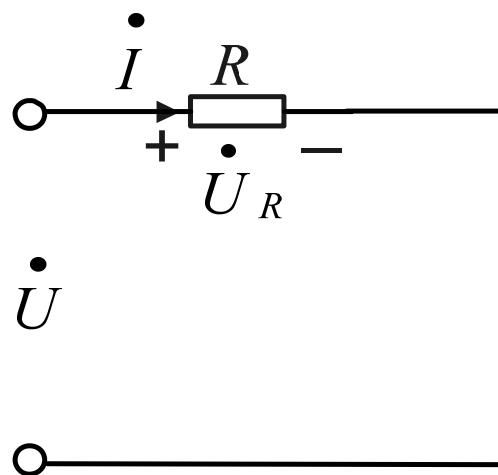


电流流入端对应的同名端为互感电压的高电位

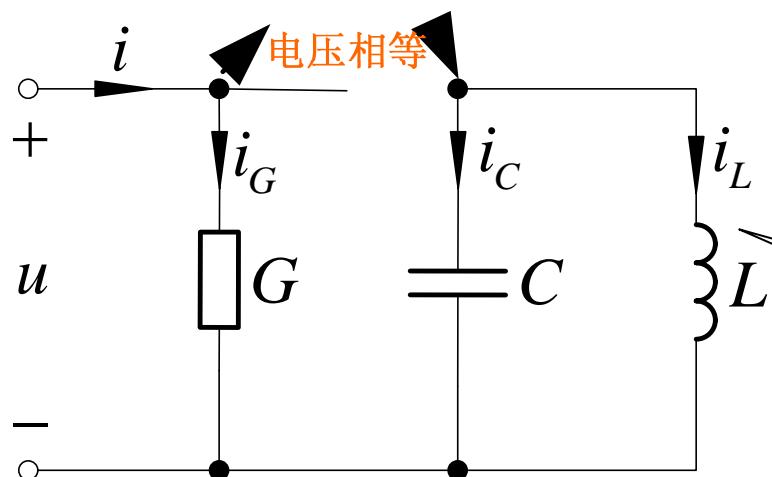
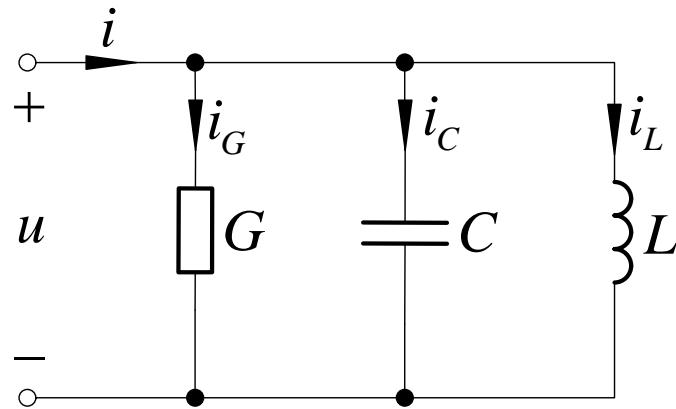


**RLC串联电路**

电容电感串联谐振对外  
等效为短路



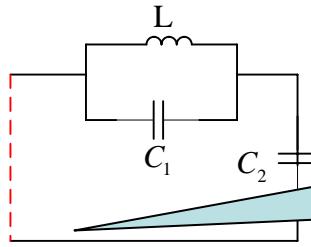
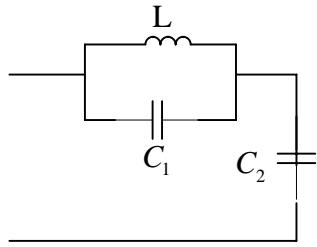
**RLC串联等效  
电路**



电容电感并联谐振对外  
等效为开路

并联等效电路

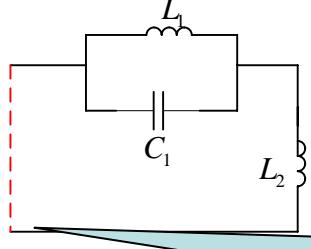
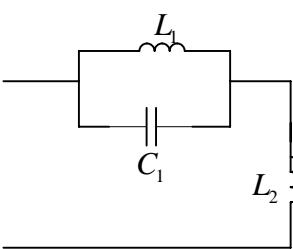
# 纯电容电感构成谐振电路的谐振频率计算



串联谐振  
对外等效  
为短路

$$\omega_1 = \frac{1}{\sqrt{LC_1}}$$

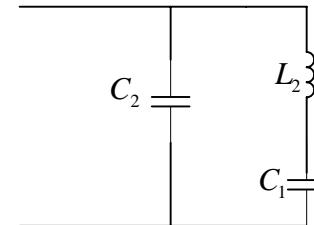
$$\omega_2 = \frac{1}{\sqrt{L(C_1+C_2)}}$$



串联谐振  
对外等效  
为短路

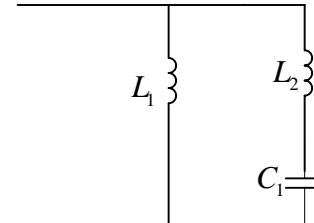
$$\omega_1 = \frac{1}{\sqrt{L_1C_1}}$$

$$\omega_2 = \frac{1}{\sqrt{C \frac{L_1L_2}{(L_1+L_2)}}}$$



并联谐振对  
外等效为开  
路

$$\omega_1 = \frac{1}{\sqrt{L_2C_1}} \quad \omega_2 = \frac{1}{\sqrt{L \frac{C_1C_2}{(C_1+C_2)}}}$$



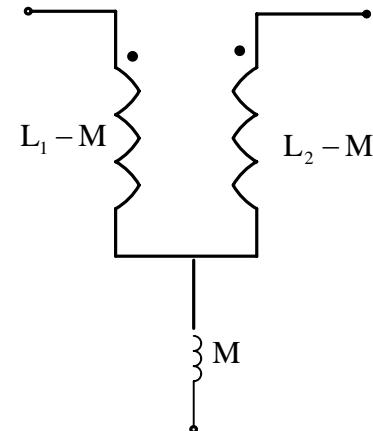
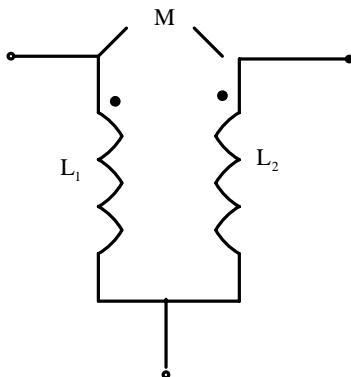
并联谐振对  
外等效为开  
路

$$\omega_1 = \frac{1}{\sqrt{L_2C_1}} \quad \omega_2 = \frac{1}{\sqrt{C_1(L_1+L_2)}}$$

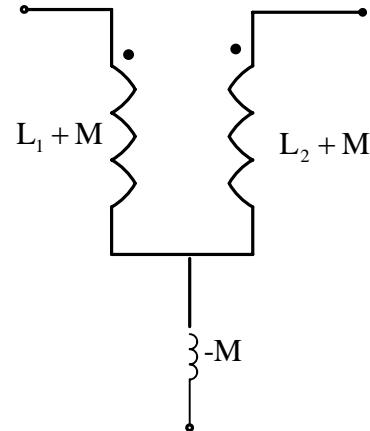
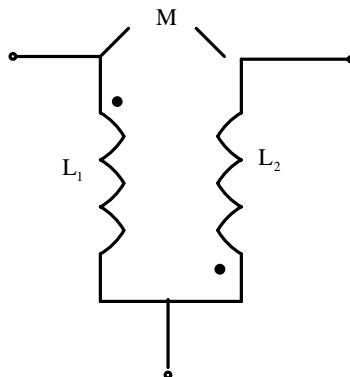
### 三、耦合电感等效电路

去耦合等效

同侧并



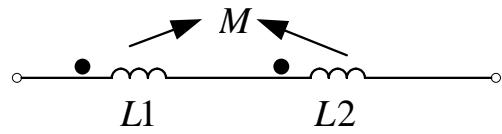
异侧并



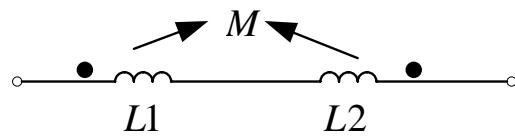
# 四、耦合电感等效电路

去耦合等效

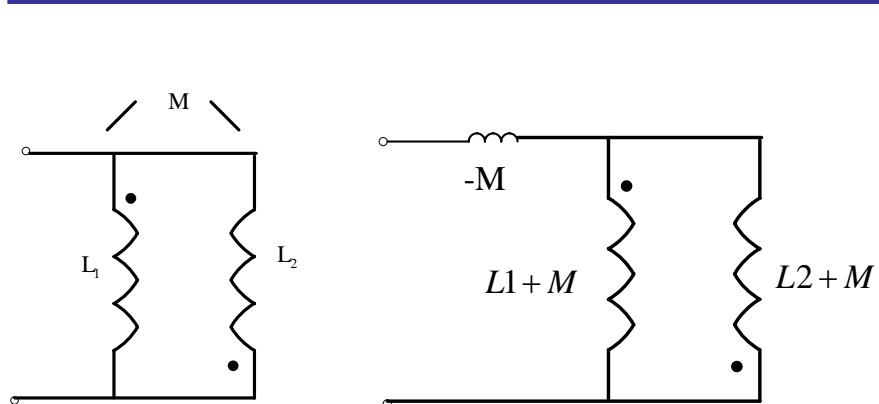
电感串并联等效电感



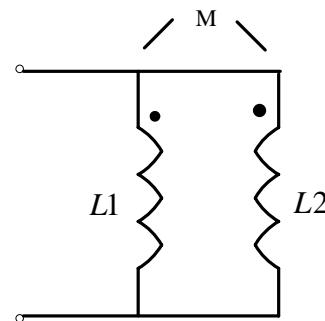
顺接



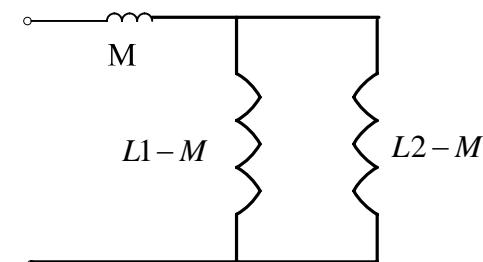
反接



异侧并

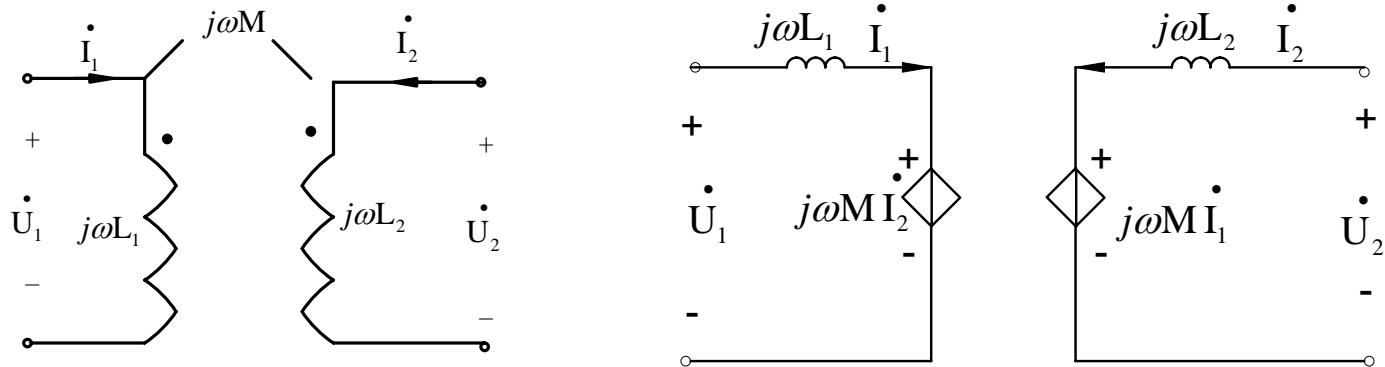


同侧并

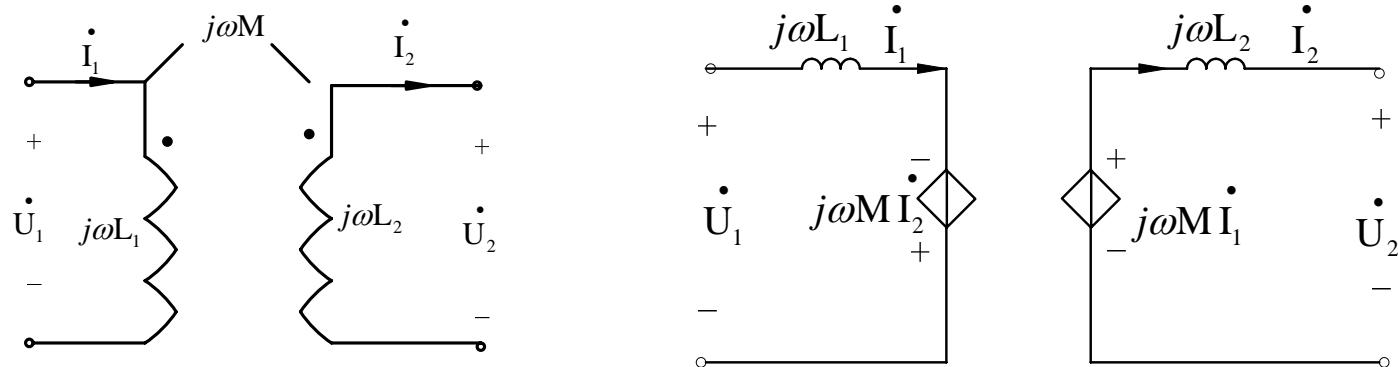


# 五、耦合电感等效电路

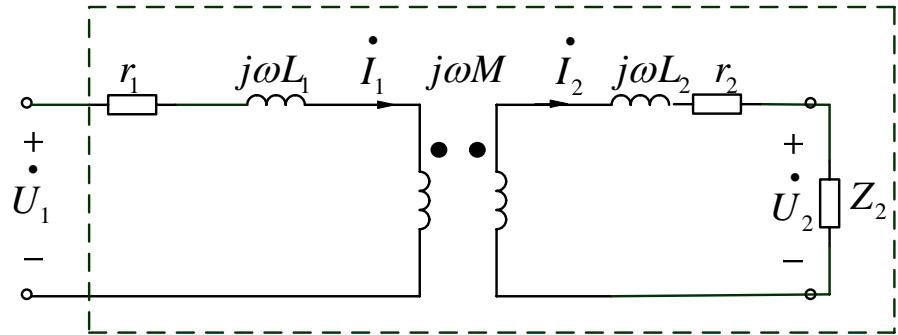
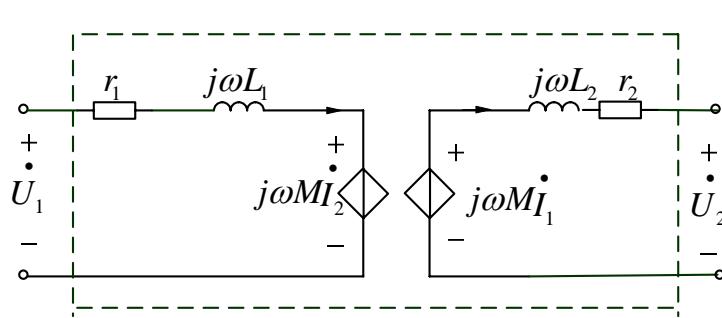
## 受控电压源等效



受控电压源的参考极性与控制量和同名端相关



# 六、空心变压器（如何识别？）



$$Z_{11} = r_1 + j\omega L_1$$

$$Z_{ref} = \frac{(\omega M)^2}{Z_{22}}$$

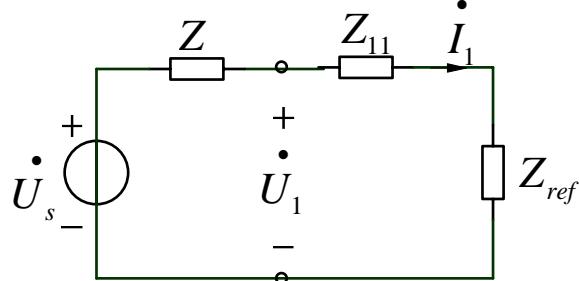
$$Z_{22} = r_2 + j\omega L_2 + Z_2$$

反映  
阻抗

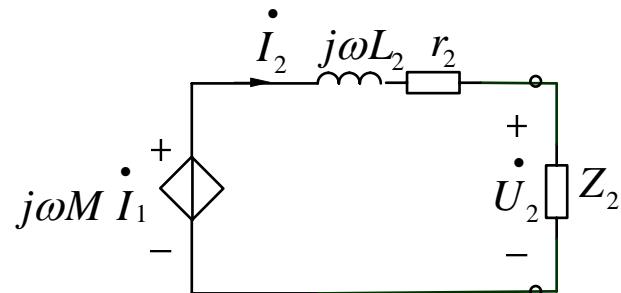
➤计算反射阻抗

➤根据原边等效电路计算电流电压

➤根据副边等效电路计算

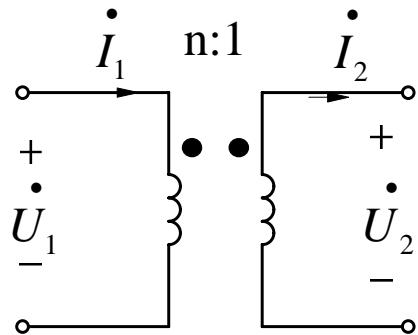


原边等效电路



副边等效电路

# 七、理想变压器

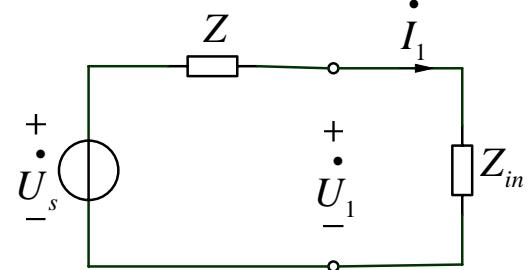
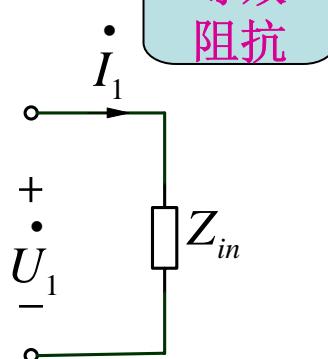
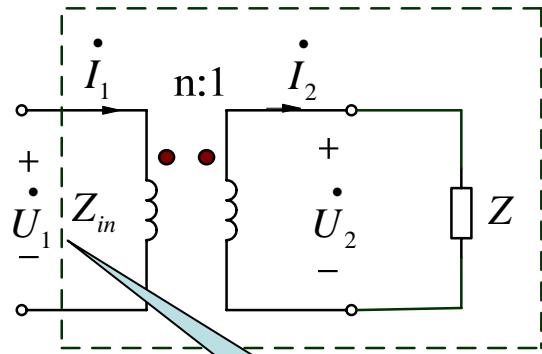


$$\begin{cases} \dot{U}_1 = n \dot{U}_2 \\ \dot{I}_2 = n \dot{I}_1 \end{cases}$$

伏安  
关系

注意伏安关系的正负

电流为零或电压为零！！！



计算等效阻抗

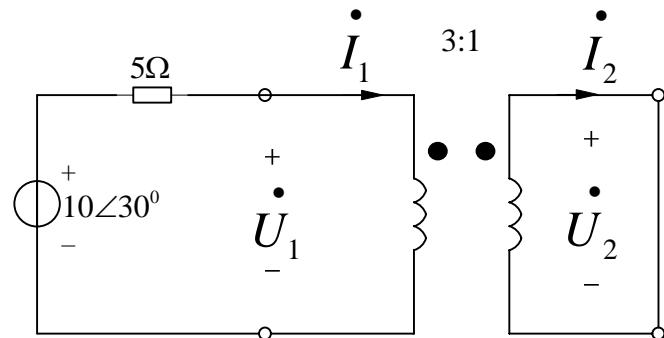
$$Z_{in} = n^2 Z$$

原边等效电路

根据原边等效电路计算电流，根据理想变压器伏安关系计算副边电压电流。

# 七、理想变压器

副边短路

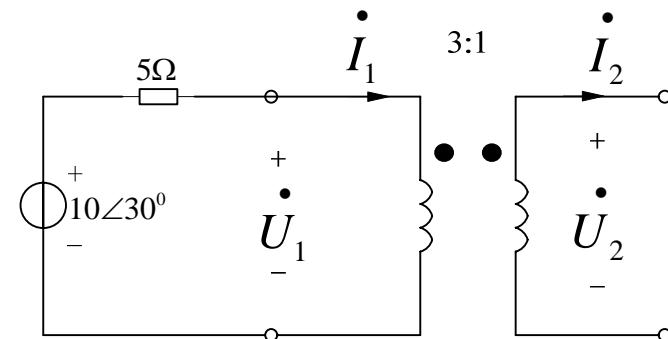


$$\dot{U}_2 = 0 \quad \dot{U}_1 = 3\dot{U}_2 = 0$$

$$\dot{I}_1 = \frac{10\angle 30^\circ}{5} = 2\angle 30^\circ$$

$$\dot{I}_2 = 3\dot{I}_1 = 6\angle -30^\circ$$

副边开路



$$\dot{I}_2 = 0 \quad \dot{I}_1 = 0$$

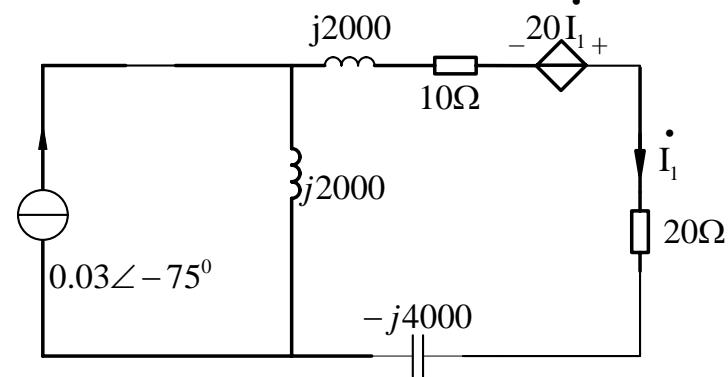
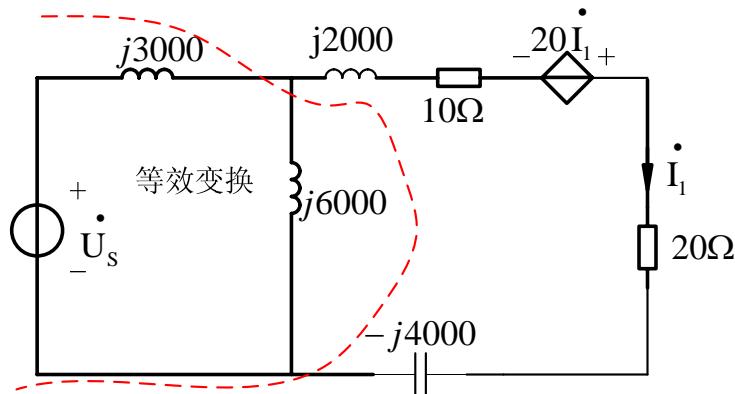
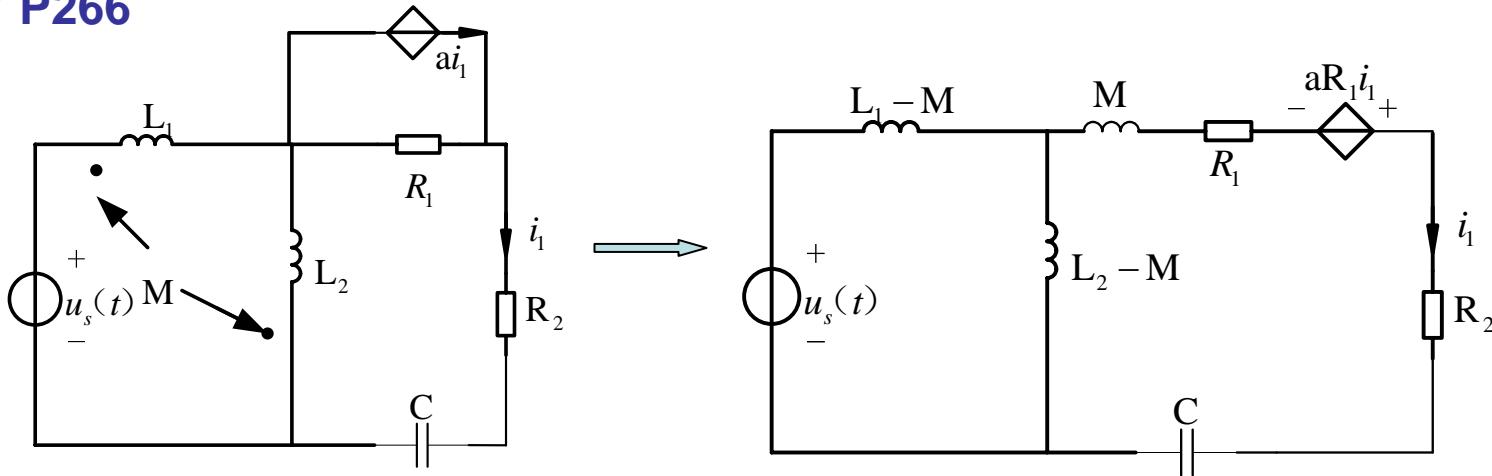
$$\dot{U}_1 = 10\angle 30^\circ$$

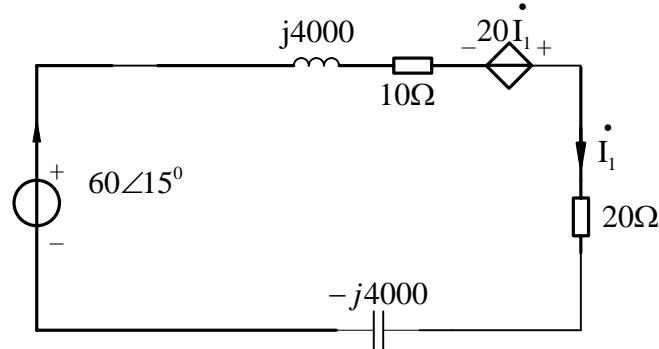
$$\dot{U}_1 = 3\dot{U}_2 \Rightarrow \dot{U}_2 = \frac{1}{3}10\angle 30^\circ$$

# 八、耦合电感问题分析

一般耦合电感电路，进行去耦等效后进行计算

9-17 P266



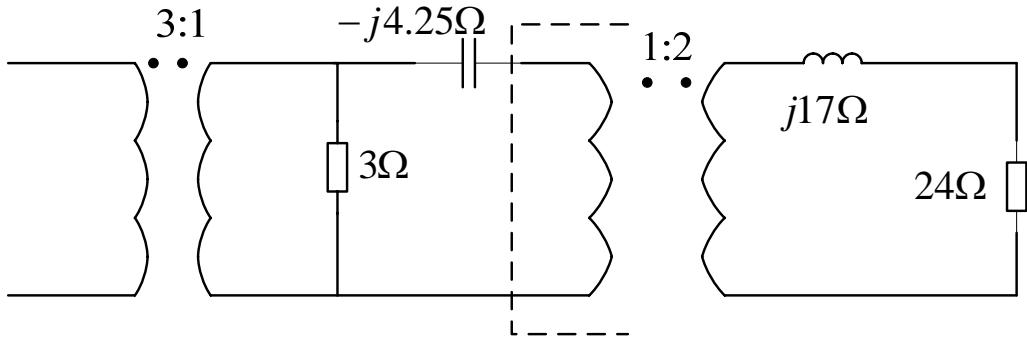


$$\dot{I}_1 = \frac{60\angle 15^0}{30 - 20} = 6\angle 15^0 A$$

$$P_{20i} = 6 \times 20 \times 6 = 720W \quad P_R = 36 \times 30 = 1080W$$

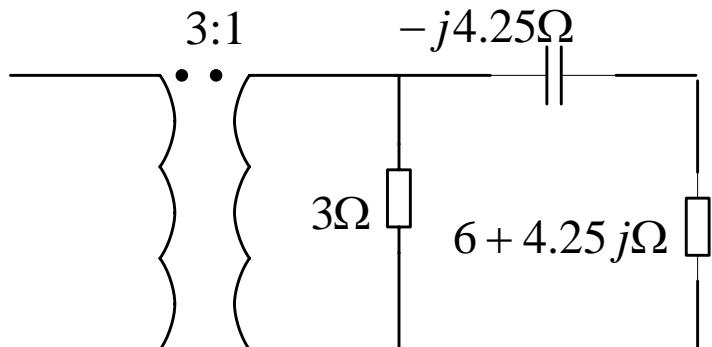
$$P_{us} = 1080 - 720 = 360W$$

求端口的输入阻抗



解

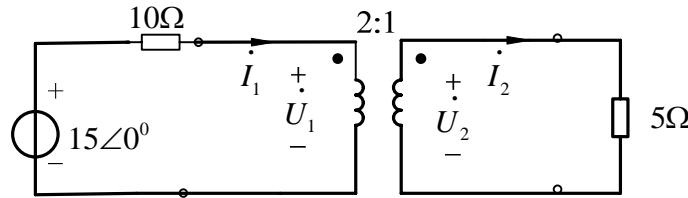
$$Z_{eq1} = \left(\frac{1}{2}\right)^2 \times (24 + 17j) = 6 + 4.25j$$



$$Z_{eq2} = \frac{(6 + 4.25j - 4.25j) \times 3}{(6 + 4.25j - 4.25j) + 3} = 2\Omega$$

$$Z_{eq} = 3^2 \times Z_{eq2} = 9 \times 2 = 18\Omega$$

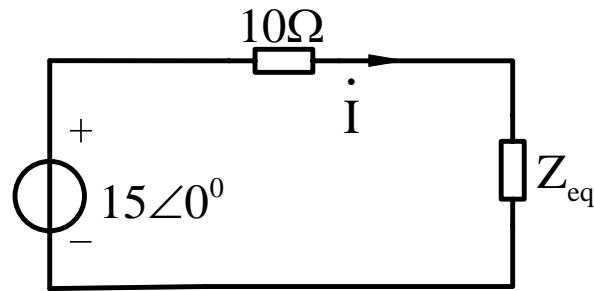
求电阻功率。



理想变压器副边开路  
与短路？？？

1) 计算等效阻抗  $Z_{eq} = 2^2 \times 5 = 20\Omega$

2) 变压器原边等效电路



$$I = \frac{15}{20 + 10} = 0.5A$$

功率  
守恒

$$P = I^2 \times R = 0.5^2 \times 20 = 5W$$

$$\text{或 } P = I_2^2 \times R = (0.5 \times 2)^2 \times 5 = 5W$$

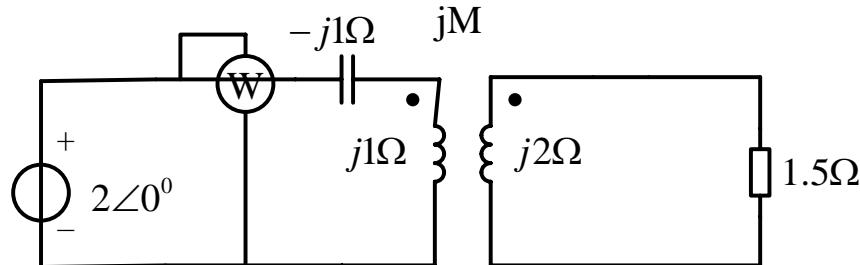
$$\begin{cases} \dot{U}_1 = 2\dot{U}_2 \\ \dot{I}_2 = 2\dot{I}_1 \end{cases}$$

双口网络参  
数方程

$$\begin{cases} 10\dot{I}_1 + \dot{U}_1 = 15\angle 0^\circ \\ \dot{U}_2 = 5\dot{I}_2 \end{cases} \quad \begin{aligned} \dot{I}_2 &= 1\angle 0^\circ A \\ P &= I^2 \times R = 1^2 \times 5 = 5W \end{aligned}$$

端口伏安  
关系

## 9-21 求互感系数M



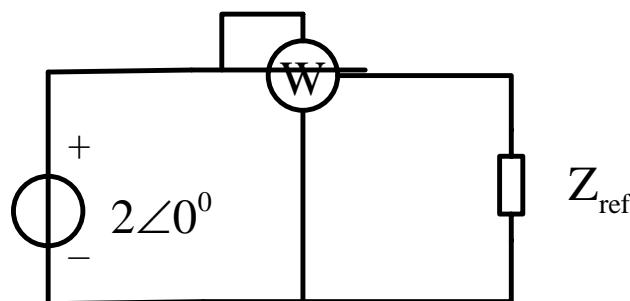
变压器副边开路  
与短路？？？

1) 计算原边和副边自阻抗  $Z_{22} = 1.5 + 2j$      $Z_{11} = 1j - 1j = 0$

2) 计算反映阻抗  $Z_{\text{ref}} = \frac{(\omega M)^2}{Z_{22}} = (\omega M)^2 \times \left( \frac{3-4j}{5} \right)$

$$\operatorname{tg} \theta = -\frac{4}{3} \quad \theta = -53^0$$

3) 变压器等效电路



$$P = 2 \times I \times \cos(-53) = 24W$$

$$I = \frac{24}{2 \times 0.6} = 20A$$

$$P = I^2 \times R = I^2 \times (\omega M)^2 \times \frac{3}{5} = 24W$$

$$M = 50mH$$

# 第八章 正弦稳态电路的相量分析

## 一、复数

$$A = a + bj = |A| \angle \theta = |A|(\cos \theta + j \sin \theta)$$

$$|A| = \sqrt{a^2 + b^2}$$

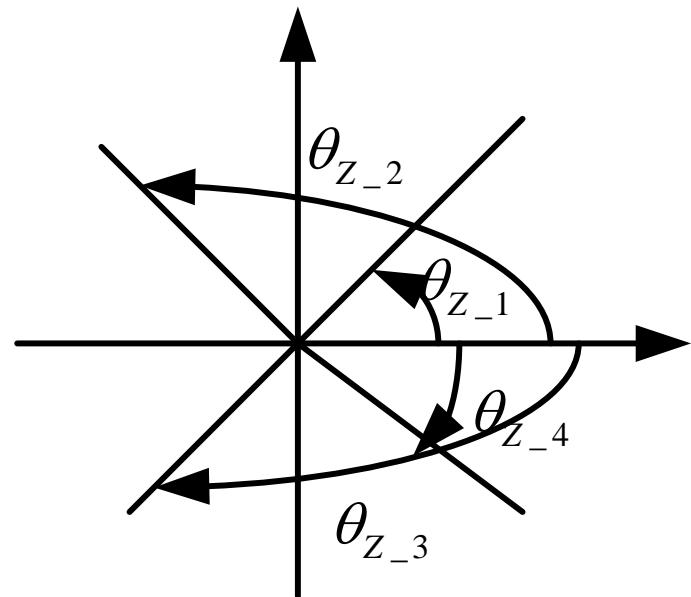
$$\theta = \arctg \frac{b}{a}$$

1和4相限

$$\theta = \pm 180^\circ + \arctg \frac{b}{a}$$

2和3相限

$$a = |A| \cos \theta \quad b = |A| \sin \theta$$



顺时针转动角度为负，逆时针转动角度为正

# 第八章 正弦稳态电路的相量分析

## 二、元件电压电流关系相量形式

$$\dot{U}_R = R \times \dot{I}_R$$

$$\dot{U}_L = j\omega l \times \dot{I}_l$$

$$\dot{U}_C = -j \frac{1}{\omega C} \times \dot{I}_C$$

$$\dot{U}_Z = (R + jX) \dot{I}_Z$$

$$Z = R + jX = |Z| \angle \theta_Z \quad \theta_Z = \arctg \left( \frac{X}{R} \right)$$

$$X = |Z| \sin \theta_Z \quad R = |Z| \cos \theta_Z$$

## 三、基尔霍夫相量形式

$$\sum \dot{U}_k = 0$$

$$\sum \dot{I}_k = 0$$

## 四、两类约束的相量图表示

### 相量变换

复数的模为正弦电压或电流的有效值或幅值

复数的辐角为正弦量的初相位

阻抗三角形  
电压三角形  
电流三角形  
功率三角形

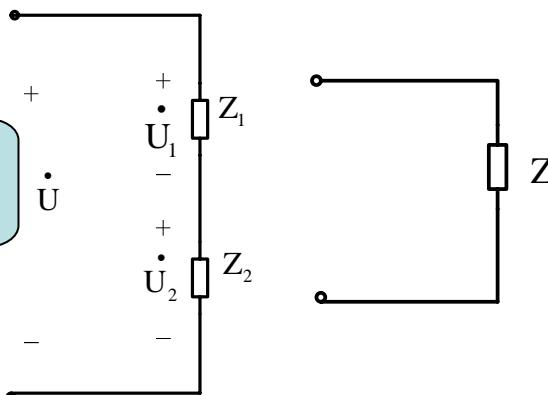
## 五、阻抗串并联

$$Z = Z_1 + Z_2$$

$$\dot{U}_2 = \frac{Z_2}{Z_1 + Z_2} \dot{U}$$

$$\dot{U}_1 = \frac{Z_1}{Z_1 + Z_2} \dot{U}$$

分压  
公式



最简单阻抗----  
单一原件

不同性质的阻抗串联，  
阻抗模变小。

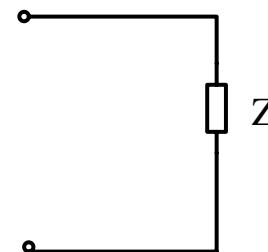
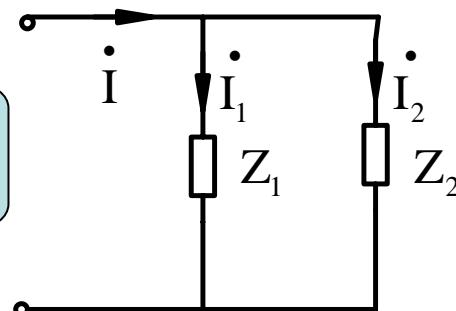
同性质阻抗串联阻抗  
模可以变大，也可以变小。

$$Z = \frac{Z_1 Z_2}{Z_1 + Z_2}$$

$$\dot{I}_1 = \frac{Z_2}{Z_1 + Z_2} \dot{I}$$

$$\dot{I}_2 = \frac{Z_1}{Z_1 + Z_2} \dot{I}$$

分流  
公式



# 六、功率

$$P = UI \cos(\theta_u - \theta_i) = UI \lambda$$

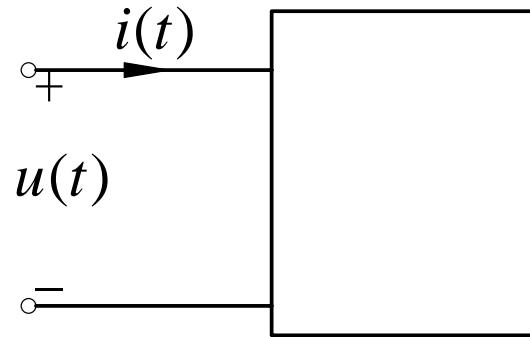
功率  
因数

$$Q = UI \sin(\theta_u - \theta_i)$$

$$\begin{cases} S = UI \\ P = S \cos \theta \\ Q = S \sin \theta \end{cases}$$

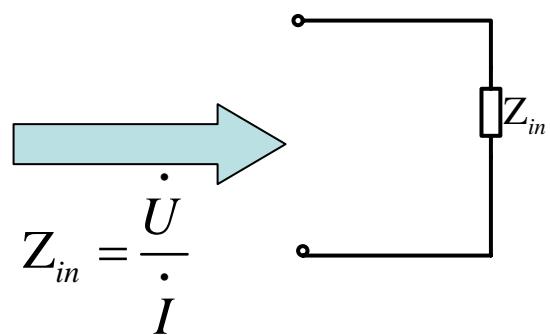
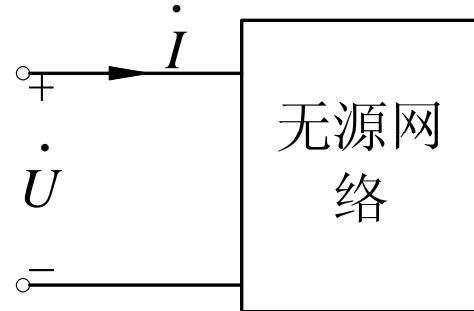
功率因数  
角

$$\tilde{S} = \dot{\bar{U}} \dot{\bar{I}}^* = UI \cos \theta + jUI \sin \theta = P + Qj$$



无源网络  
等效阻抗

列写端口电压电  
流相量方程



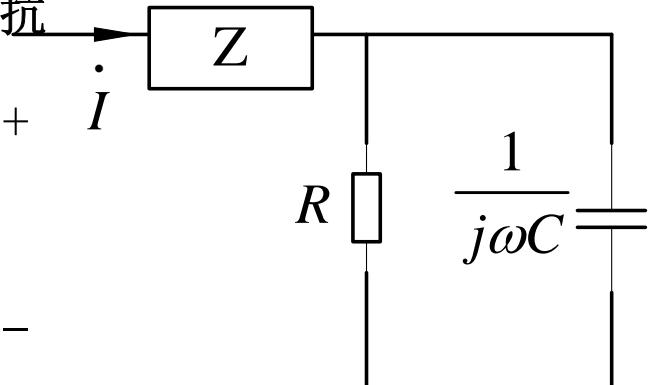
已知电源电压  $\dot{U} = 165\angle 0^\circ$  电源提供的复功率为  $\tilde{S} = 1650 - j825$

负载Z吸收的复功率为  $\tilde{S}_1 = 1250 - j625$  求电阻和容抗

$$\text{解: 1) } \tilde{S} = \dot{U} \times \dot{I}^* \Rightarrow 1650 - j825 = 165\angle 0^\circ \dot{I}^* + \dot{I}$$

$$\dot{I}^* = \frac{1650 - j825}{165\angle 0^\circ} = 10 - j5A$$

2)



$$1250 - j625 = \dot{U}_z \dot{I}^* \quad \dot{U}_z = \frac{1250 - j625}{10 - j5} = 125\angle 0^\circ$$

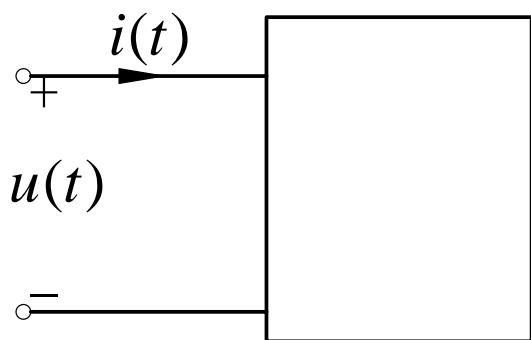
$$\dot{U}_R = \dot{U} - \dot{U}_z = 165\angle 0^\circ - 125\angle 0^\circ = 40\angle 0^\circ$$

$$1650 - j825 = 1250 - j625 + \tilde{S}_2$$

$$\tilde{S}_2 = 400 - j200 \quad P_R = 400 \quad Q_C = 200$$

$$P_R = \frac{U_R^2}{R} \Rightarrow R = \frac{400}{1600} = 0.25\Omega \quad Q_C = U_R^2 \omega C \Rightarrow \omega C = \frac{200}{1600} = 0.125$$

## 8-1 求串联最简等效电路及参数



$$u = 10\sqrt{2} \sin(t + 10^\circ)$$

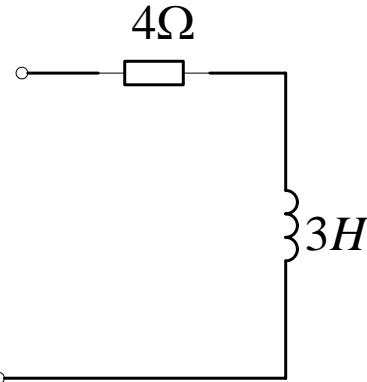
$$i = 2\sqrt{2} \cos(t - 117^\circ)$$

$$\dot{U} = 10\angle 10^\circ V$$

$$\dot{I} = 2\angle -27^\circ A$$

$$Z = \frac{\dot{U}}{\dot{I}} = \frac{10\angle 10^\circ}{2\angle -27^\circ} = 4 + j3\Omega$$

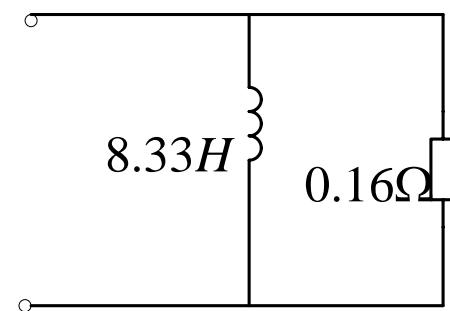
$$R = 4\Omega \quad \omega L = 3\Omega \quad L = 3H$$



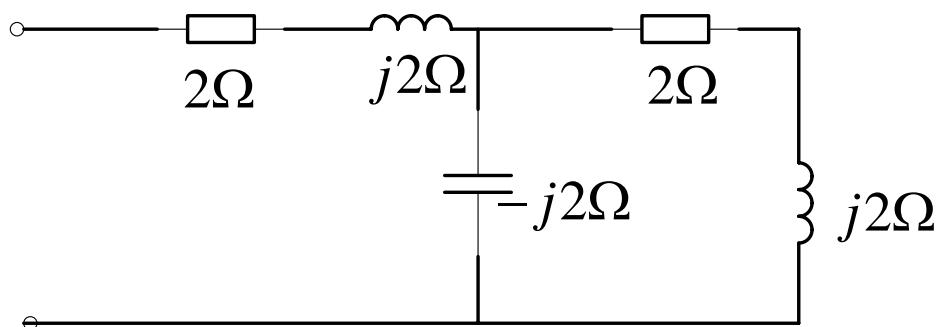
$$Y = \frac{\dot{I}}{\dot{U}} = \frac{2\angle -27^\circ}{10\angle 10^\circ} = 0.16 - 0.12jS$$

$$G = 0.16S$$

$$\frac{1}{\omega L} = 0.12 \quad L = 8.33H$$



## 8-1 求等效阻抗



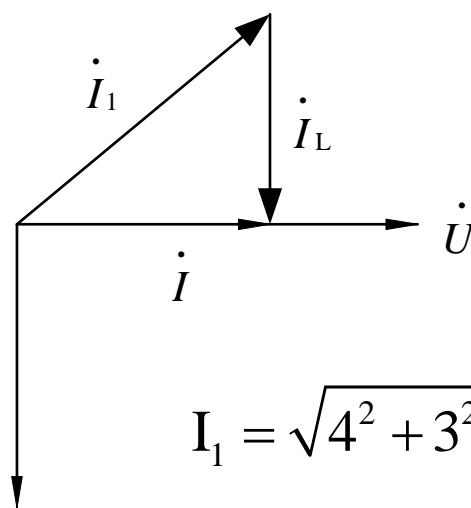
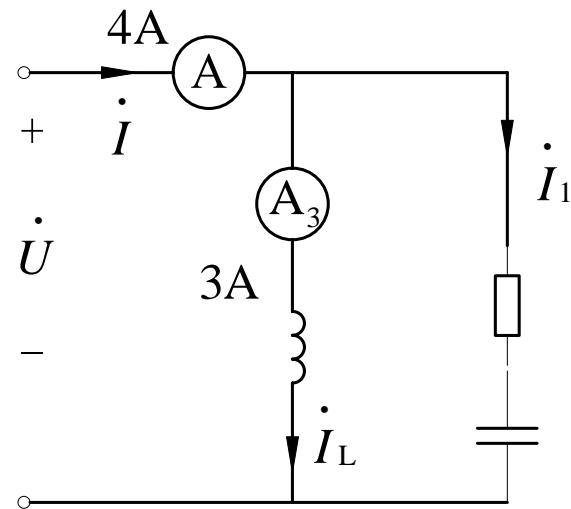
$$Z_3 = 2 + j2\Omega \quad Z_2 = -j2\Omega$$

$$Z_1 = 2 + j2\Omega$$

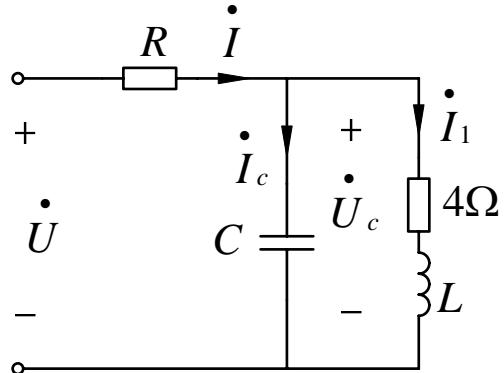
$$Z_3 // Z_2 = \frac{Z_2 \times Z_3}{Z_3 + Z_2} = \frac{-j2 \times (2 + j2)}{-j2 + 2 + j2} = 2 - j2$$

$$Z = Z_3 // Z_2 + Z_1 = 2 - j2 + 2 + j2 = 4\Omega$$

已知电压电流同相位，求电流  $I_1$



$$I_1 = \sqrt{4^2 + 3^2} = 5A$$



$$I_1 = 5A \quad I_c = 3A \quad U = 65V \quad \text{端口电压电流同相}$$

求R,L,C ,w=3000rad。

解 令电容电压为参考，则  $\dot{U}_c = U_c \angle 0^\circ V$

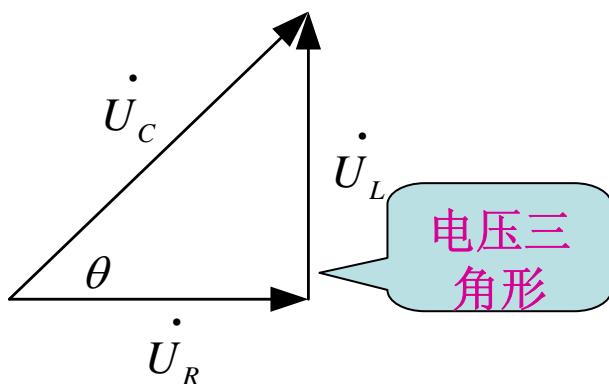
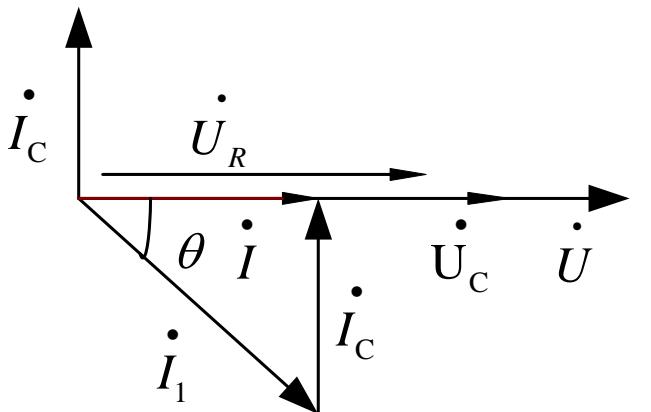
$$I = \sqrt{I_1^2 - I_c^2} = \sqrt{25 - 9} = 4A \quad \theta = \arctg \frac{3}{4} = 37^\circ$$

$$\frac{\omega L}{4} = \tan 37^\circ \quad \omega L = \tan 37^\circ \times 4 = 3 \quad \omega = 1mH$$

$$U_{4\Omega} = 4 \times 5 = 20V \quad U_c = \frac{20}{\cos 37^\circ} = 25V$$

$$U_R = 65 - 25 = 40V \quad R = \frac{40}{4} = 10\Omega$$

$$\frac{1}{\omega C} = \frac{40}{3} \quad C = \frac{3}{3000 \times 25} = 40\mu F$$

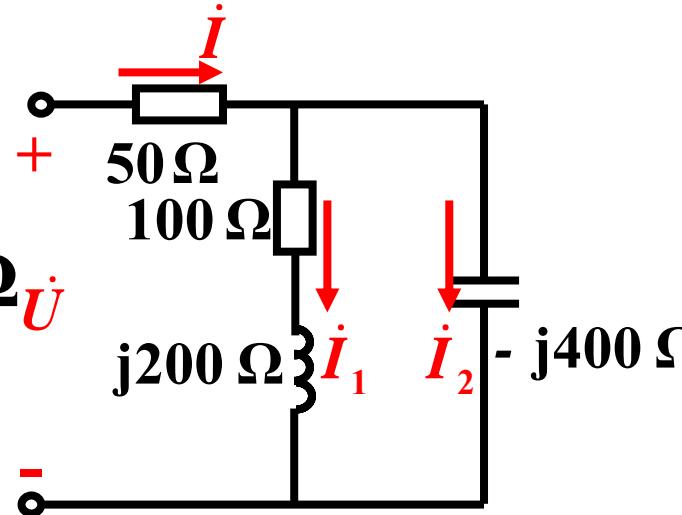


计算图中所示的电流

$$\dot{U} = 220 \angle 0^\circ \text{ V}$$

$$Z_1 = R_1 + jX_L = (100 + j1200) \Omega$$

$$Z_2 = -jX_C = -j140 \Omega$$



$$Z = [50 + \frac{(100 + j200)(-j400)}{100 + j200 - j400}] = (50 + 320 + j240) \\ = 440 \angle 33^\circ \Omega$$

$$\dot{I} = \frac{\dot{U}}{Z} = \frac{220 \angle 0^\circ}{440 \angle 33^\circ} \text{ A} = 0.5 \angle -33^\circ \text{ A}$$

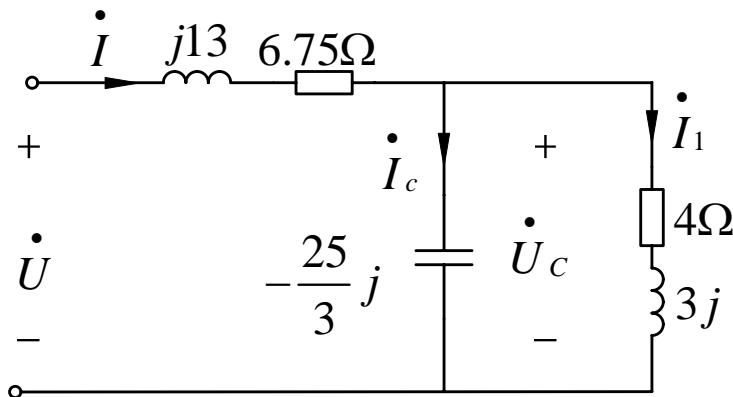
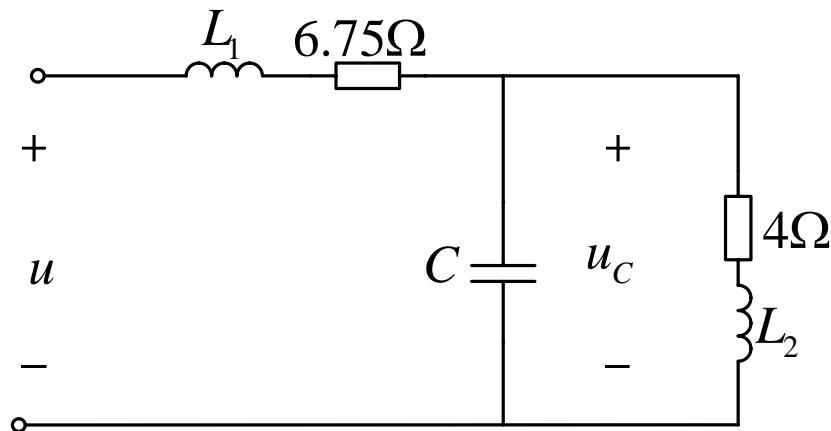
$$\dot{I}_1 = \frac{Z_2}{Z_1 + Z_2} \dot{I} = \frac{-j400}{100 + j200 - j400} \times 0.5 \angle -33^\circ \text{ A} \\ = 0.89 \angle -59.6^\circ \text{ A}$$

已知  $U_c = 50V$ ,  $L_1 = 6.5H$ ,  $L_2 = 1.5H$ , 电源角频率为  $2rad/s$ ,  $C = 0.06F$ , 求端口电压和端口电流的有效值并计算电路吸收的有功功率, 无功功率, 复功率。

### (1) 画相量模型

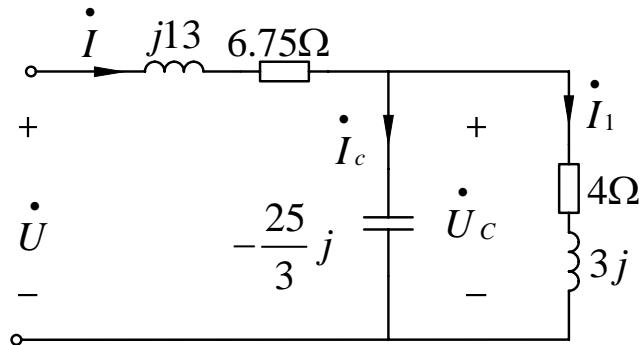
$$\omega L_1 = 2 \times 6.5 = 13\Omega \quad \omega L_2 = 2 \times 1.5 = 3\Omega$$

$$\frac{1}{\omega C} = \frac{1}{2 \times 0.06} = \frac{25}{3}\Omega$$



### (2) 令电容电压为参考相量

$$\dot{U}_C = 50\angle 0^{\circ}V$$



$$\dot{I}_c = \frac{\dot{U}_c}{-25/3j} = \frac{50\angle 0^\circ}{25/3\angle -90^\circ} = 6\angle 90^\circ$$

$$\dot{I}_1 = \frac{\dot{U}_c}{4+3j} = \frac{50\angle 0^\circ}{5\angle 37^\circ} = 10\angle -37^\circ$$

$$\begin{aligned}\dot{I} &= \dot{I}_1 + \dot{I}_c = 10\angle -37^\circ + 6\angle 90^\circ \\ &= 10(\cos -37 + j \sin -37) + 6j \\ &= 10 \times 0.8 - 10 \times 0.6j + 6j = 8\end{aligned}$$

$$\begin{aligned}\dot{U} &= \dot{I}(6.75 + 13j) + \dot{U}_c \\ &= 54 + 104j + 50 \\ &= 104 + 104j = 104\sqrt{2}\angle 45^\circ\end{aligned}$$

端口电压  
电流相位差

$$(3) \quad \theta = 45 - 0 = 45^\circ$$

$$\lambda = \cos 45^\circ = 0.707$$

$$P = UI \cos \theta = 104 \times 8 \times \sqrt{2}/2 = 416\sqrt{2}W$$

$$Q = UI \sin \theta = 104 \times 8 \times \sqrt{2}/2 = 416\sqrt{2}VAr$$

$$S = UI = 832VA$$

$$\begin{aligned}\tilde{S} &= \tilde{U} \overset{*}{\tilde{I}} = 104\sqrt{2}\angle 45^\circ \times 8 \\ &= 416\sqrt{2} + 416\sqrt{2}j\end{aligned}$$

列写图所示电路节点电压方程

已知  $u_{s1} = 5\sqrt{2} \sin(100t + 30^\circ)$   $u_{s2} = 10\sqrt{2} \sin(100t + 60^\circ)$

$$i_s = 3\sqrt{2} \cos(100t - 60^\circ)$$

解：1) 求电路参数

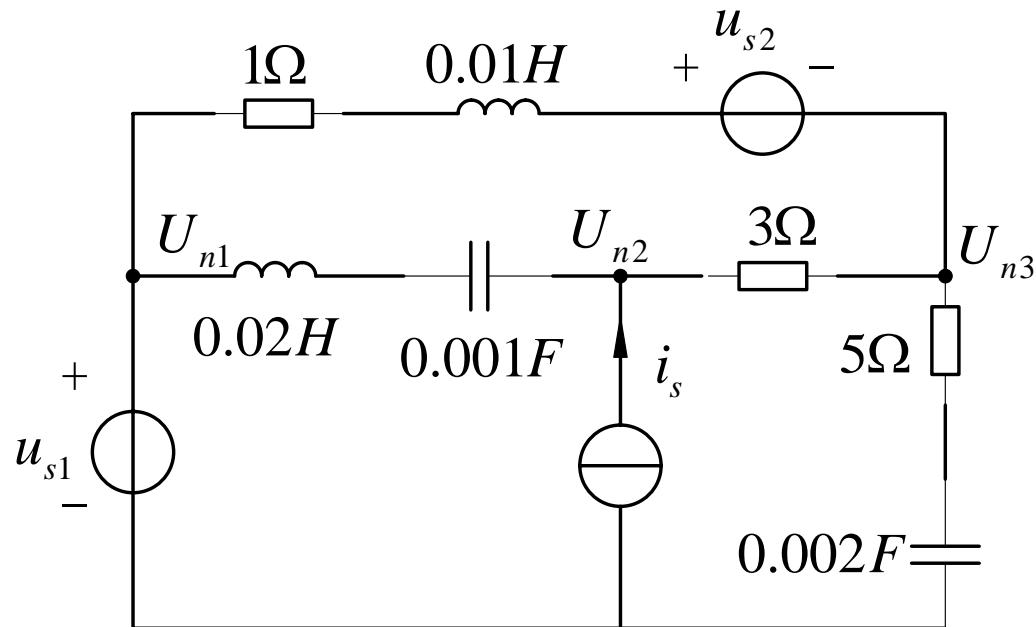
$$\omega L_1 = 100 \times 0.01 = 1\Omega$$

$$\omega L_2 = 100 \times 0.02 = 2\Omega$$

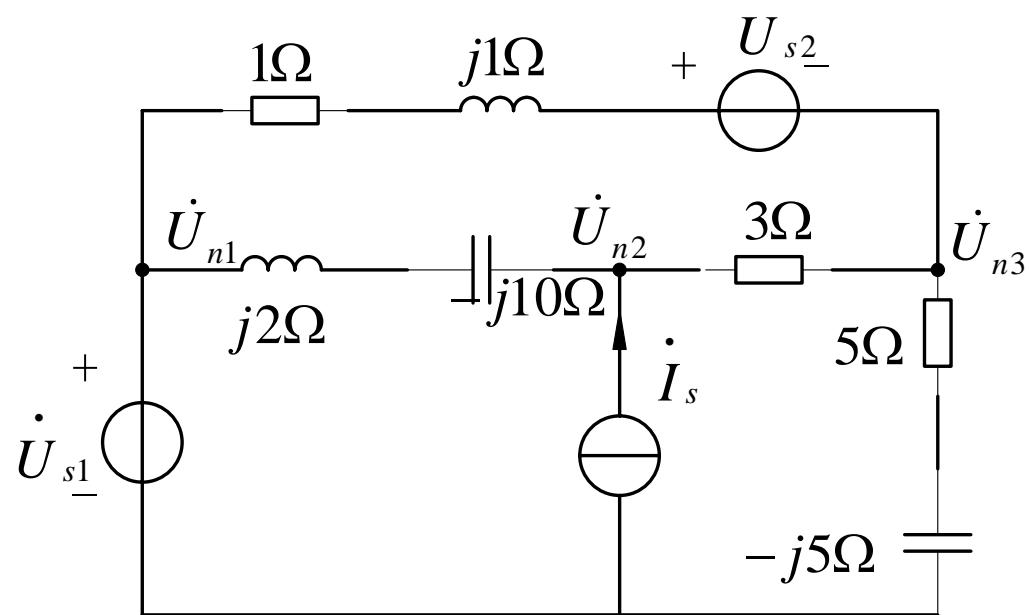
$$\frac{1}{\omega C_1} = \frac{1}{100 \times 0.001} = 10\Omega$$

$$\frac{1}{\omega C_2} = \frac{1}{100 \times 0.002} = 5\Omega$$

$$\dot{U}_{s1} = 5\angle 30^\circ \quad \dot{U}_{s2} = 10\angle 60^\circ \quad \dot{i}_s = 3\angle 30^\circ$$



2) 画相量模型



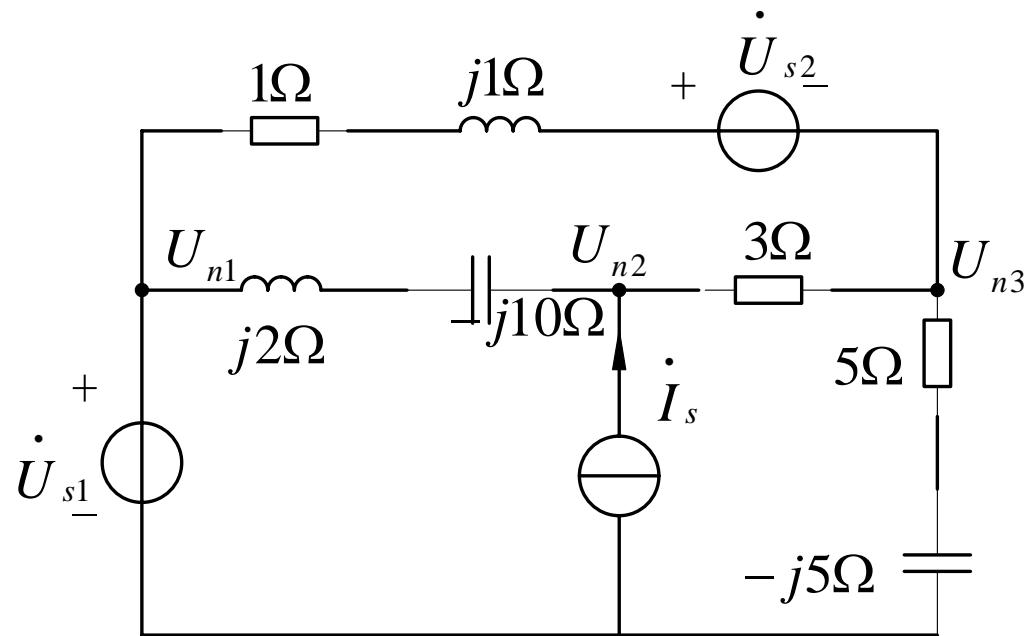
3) 列写节点1方程

$$\dot{U}_{n1} = \dot{U}_{s1} = 5\angle 30^\circ$$

4) 列写节点2方程

$$\left(\frac{1}{j2 - j10} + \frac{1}{3}\right)\dot{U}_{n2} - \frac{1}{3}\dot{U}_{n3} - \left(\frac{1}{j2 - j10}\right)\dot{U}_{n1} = 3\angle 30^\circ$$

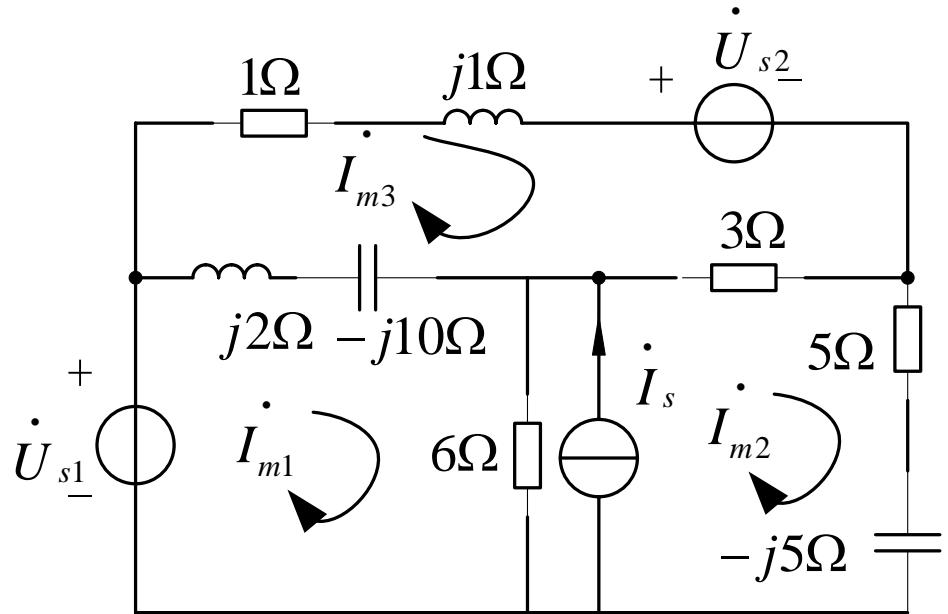
$$\left(\frac{1}{-j8} + \frac{1}{3}\right)\dot{U}_{n2} - \frac{1}{3}\dot{U}_{n3} + \frac{1}{j8}\dot{U}_{n1} = 3\angle 30^\circ$$



5) 列写节点3方程

$$\left( \frac{1}{5-j5} + \frac{1}{3} + \frac{1}{1+j} \right) \dot{U}_{n3} - \frac{1}{3} \dot{U}_{n2} - \left( \frac{1}{1+j} \right) \dot{U}_{n1} = -\frac{10\angle 60^\circ}{1+j} = -\frac{10\angle 60^\circ}{\sqrt{2}\angle 45^\circ} = -5\sqrt{2}\angle 15^\circ$$

列写图所示电路网孔电流方程

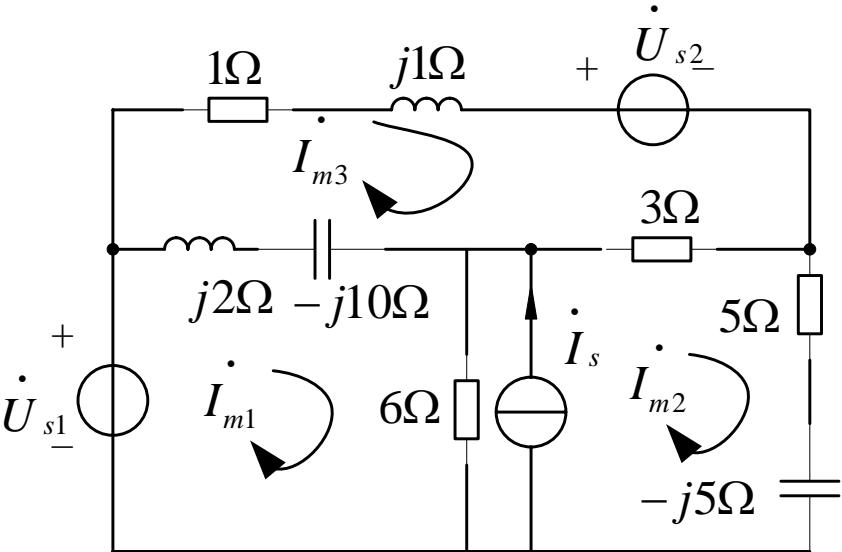


1) 列写网孔1方程

$$(j2 - j10 + 6)\dot{I}_{m1} - 6\dot{I}_{m2} - (j2 - j10)\dot{I}_{m3} = \dot{U}_{s1} - \dot{I}_3 \times 6$$

$$(-j8 + 6)\dot{I}_{m1} - 6\dot{I}_{m2} + (j8)\dot{I}_{m3} = 5\angle 30^\circ - 3\angle 30^\circ \times 6 = -13\angle 30^\circ$$

列写图所示电路网孔电流方程



2) 列写网孔2方程

$$(3 - j5 + 5)\dot{I}_{m1} - 6\dot{I}_{m1} - 3\dot{I}_{m3} = \dot{I}_3 \times 6$$

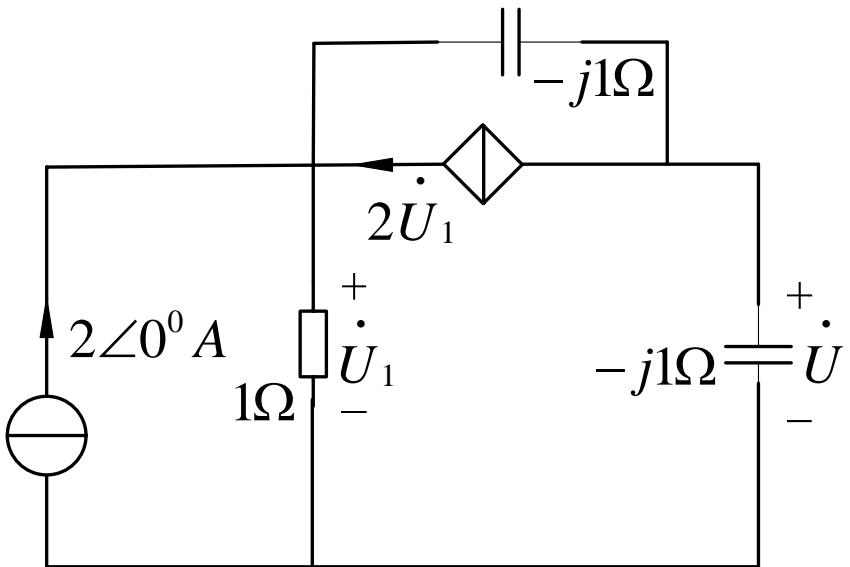
$$(8 - j5)\dot{I}_{m1} - 6\dot{I}_{m1} - 3\dot{I}_{m3} = 18\angle 30^0$$

3) 列写网孔3方程

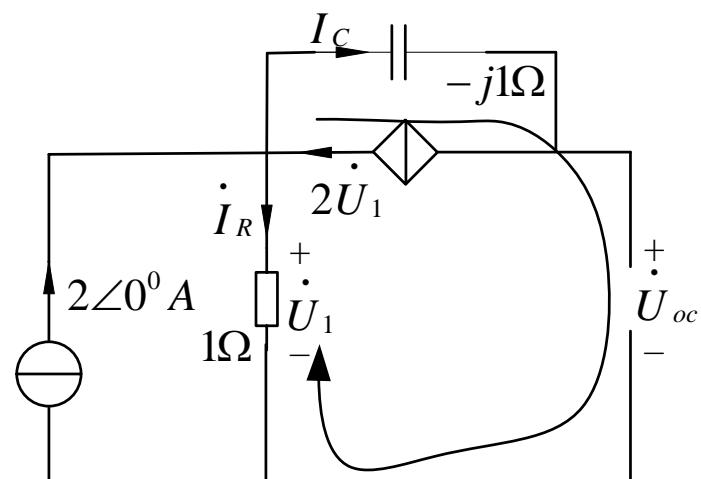
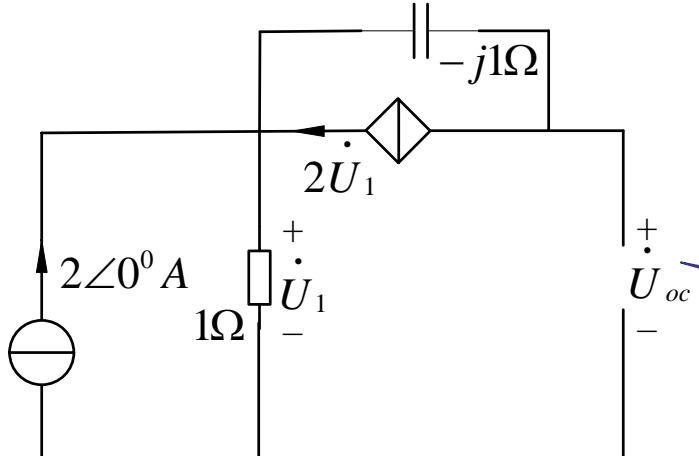
$$(3 + j2 - j10 + 1 + j1)\dot{I}_{m3} - (j2 - j10)\dot{I}_{m1} - 3\dot{I}_{m2} = -\dot{U}_{s2}$$

$$(4 - j7)\dot{I}_{m3} + j8\dot{I}_{m1} - 3\dot{I}_{m2} = -10\angle 60^0$$

电路如图所示，利用戴维南定理求电压 $U$



解：1) 求开路电压



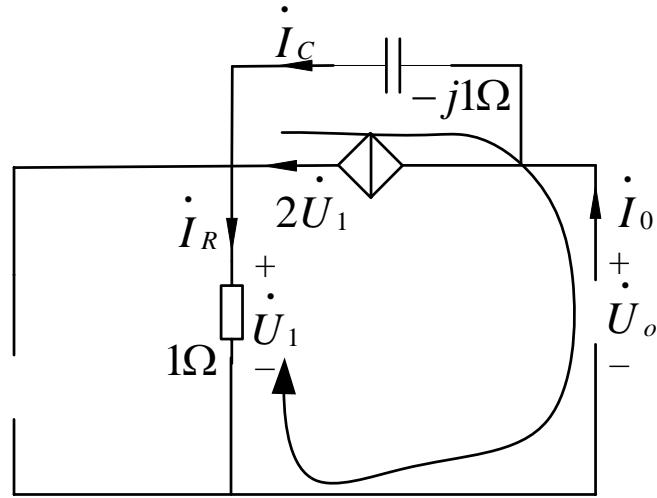
$$\dot{I}_c = 2\dot{U}_1 \quad \dot{I}_R = 2\angle 0^\circ$$

$$\dot{U}_1 = 1 \times \dot{I}_R = 2\angle 0^\circ$$

$$\dot{U}_{oc} = \dot{U}_1 - \dot{I}_c \times -j1 = 2 + j4$$

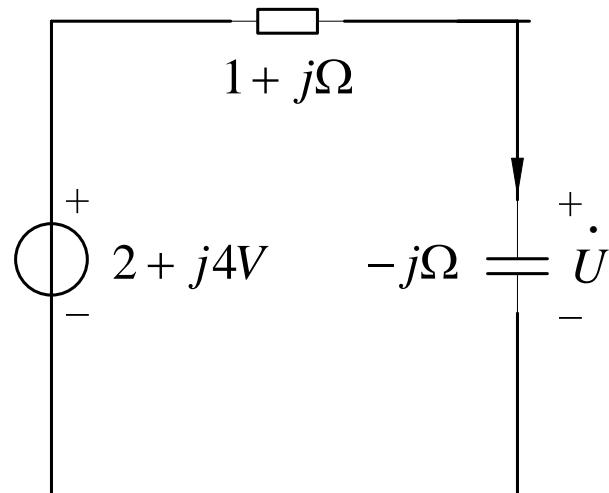
断开待求支路

## 2) 等效阻抗



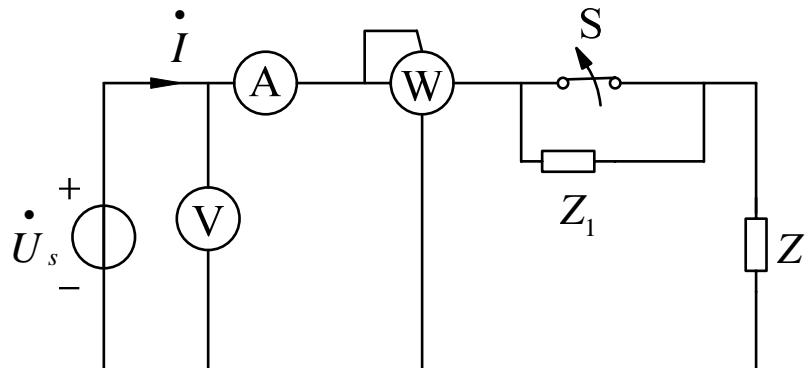
$$\begin{aligned}\dot{I}_0 &= \dot{I}_c + 2\dot{U}_1 & \dot{I}_R &= \dot{I}_0 = \frac{\dot{U}_1}{1} \\ \dot{I}_c &= -\dot{U}_1 \\ \dot{U}_0 &= \dot{I}_c \times -j1 + \dot{U}_1 = (1+j)\dot{U}_1 \\ Z_{eq} &= \frac{\dot{U}_0}{\dot{I}_0} = \frac{(1+j)\dot{U}_1}{\dot{U}_1} = 1 + j\Omega\end{aligned}$$

## 3) 画等效电路并求解



$$\begin{aligned}\dot{U} &= \frac{2 + j4}{1 + j - j} \times -j = 4 - j2 \\ &= 2\sqrt{5} \angle -26.6^\circ \text{V}\end{aligned}$$

## 8-20



已知开关闭合,  $U=220V$ ,  $P=1000W$ ,  
 $I=10A$ ,开关打开,  $U=220V$ ,  $P=1600W$ ,  
 $I=12A$ , 阻抗 $Z_1$ 为感性, 求 $Z_1$ 和 $Z$ 。

(1) 令

$$Z = R + Xj \quad Z_1 = R_1 + X_1 j$$

串联阻抗后电流变大,  
所以阻抗性质相反

$$(2) \quad P = I^2 R \quad 1000 = 10^2 R \quad R = 10\Omega \quad Z = 10 - 19.6j$$

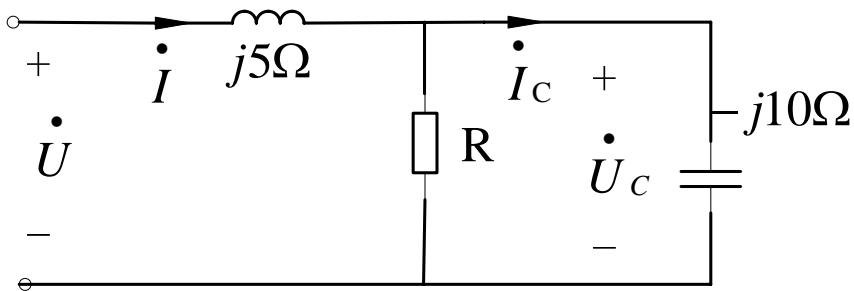
$$|Z| = \frac{U}{I} = \sqrt{R^2 + X^2} = 22 \quad X = -19.6\Omega$$

$$(3) \quad 1600 = 12^2 (R + R_1) \quad R_1 = 1.11\Omega \quad X_1 - 19.6 = \pm 14.58$$

$$|Z + Z_1| = \frac{U}{I} = \sqrt{(R + R_1)^2 + (X_1 + X)^2} = \frac{220}{12} \quad Z_1 = 1.11 + 5.02j$$

$$Z_1 = 1.11 + 34.18j$$

## 8-2



已知  $C=0.02F$ ,  $L=1H$ , 电路消耗的功率  $P=10W$ , 求电路功率因数。

$$i_c(t) = \sqrt{2} \sin(5t + 90^\circ) A$$

$$(1) \quad \omega L = 5 \times 1 = 5\Omega$$

$$\frac{1}{\omega C} = \frac{1}{5 \times 0.02} = 10\Omega$$

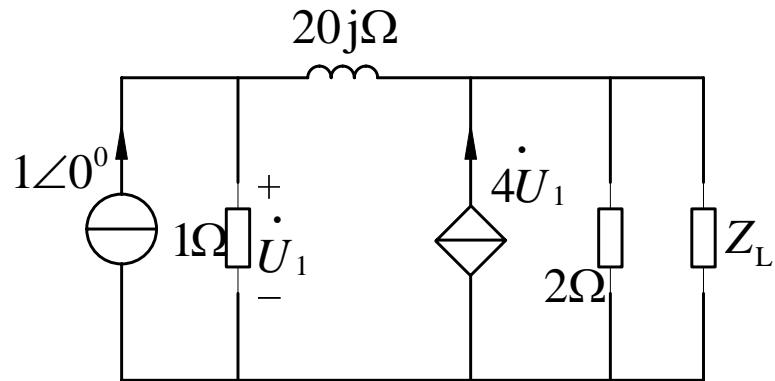
$$\dot{I}_c = 1 \angle 90^\circ A$$

$$(2) \quad \dot{U}_c = -j10 \times 1 \angle 90^\circ = 10 \angle 0^\circ V \quad P = \frac{U_c^2}{R} = 10 \quad R = \frac{100}{10} = 10\Omega$$

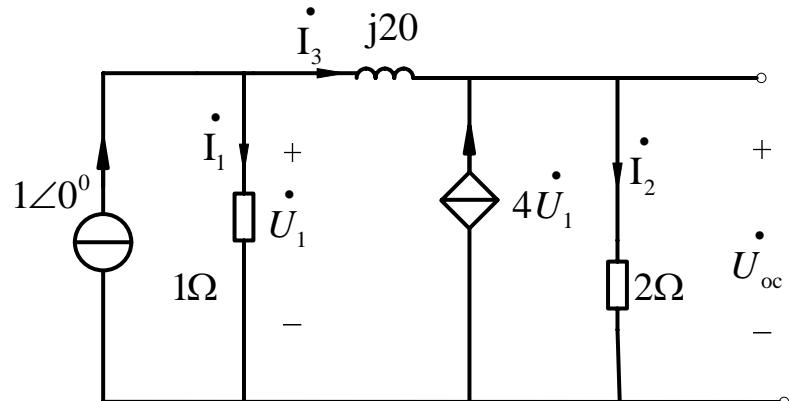
$$Z = j5 + -j10 // 10 = 5j + 5 - 5j\Omega = 5\Omega$$

$$\lambda = \cos 0^\circ = 1$$

8-34 正弦稳态电路中，负载 $Z_L$ 可以变化，问 $Z_L$ 为多少的时候其获得功率最大



解 1) 求开路电压



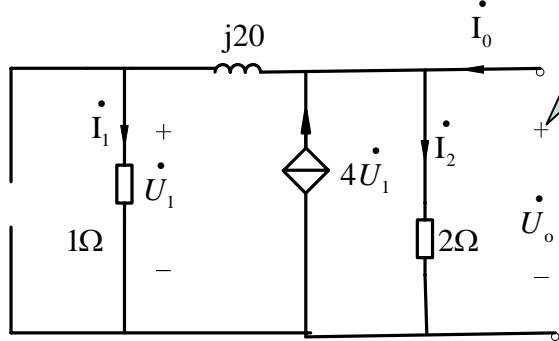
$$\dot{I}_3 = \dot{I}_2 - 4\dot{U}_1 \quad \dot{U}_1 + \dot{I}_3 = 1\angle 0^\circ$$

$$\dot{U}_1 = (1 - \dot{U}_1) \times 20j + 2 \times \dot{I}_2$$

$$\dot{U}_1 = \frac{2 + 20j}{20j - 5} = 1\angle -20^\circ V$$

$$\dot{U}_{oc} = 2 + 6\angle -20^\circ = 8\angle -15^\circ$$

## 解 2) 求等效阻抗



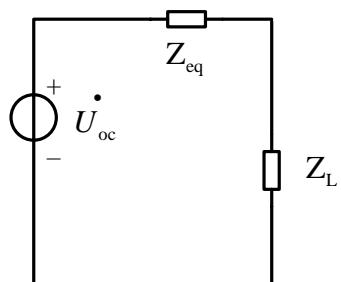
列写端口电压电流方程

$$\dot{U}_0 = \dot{U}_1 + \dot{U}_1 \times j20$$

$$\dot{I}_0 = \frac{\dot{U}_0}{2} - 4\dot{U}_1 + \dot{U}_1 = -2.5\dot{U}_1 + 10j \times \dot{U}_1$$

$$Z_{eq} = \frac{\dot{U}_0}{\dot{I}_0} = \frac{1 + j20}{-2.5 + 10j} = 1.858 - 0.565j$$

## 3) 等效电路

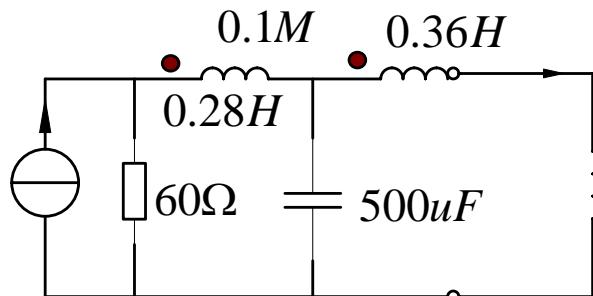


为等效阻抗的共轭复数

$$Z = Z_{eq}^* = 1.858 + 0.565j$$

$$P_{max} = \frac{U_{oc}^2}{4 \times 1.858} = 8.11W$$

已知  $i_s = 4\sqrt{2} \sin(100t) A$  求电容电压电流



(1)

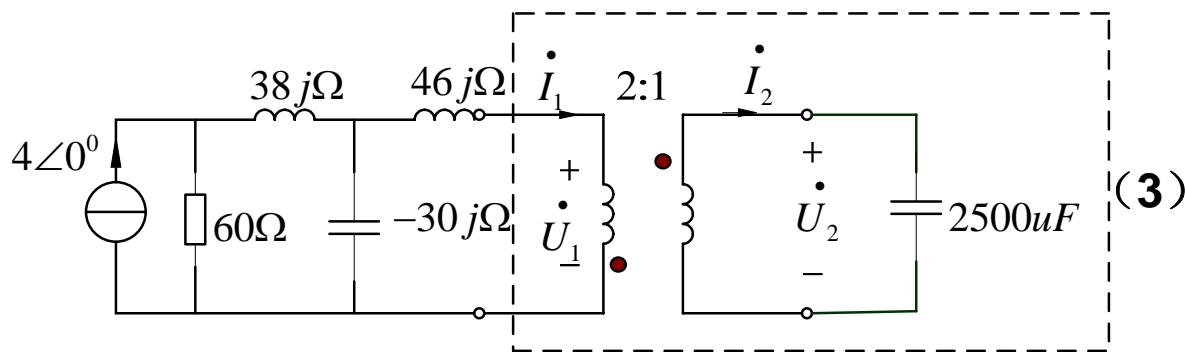
$$\omega(0.28 + 0.1) = 100 \times 0.38 = 38\Omega$$

$$\omega(0.36 + 0.1) = 100 \times 0.46 = 46\Omega$$

$$\omega M = 100 \times 0.1 = 10\Omega$$

$$\frac{1}{\omega C_1} = \frac{1}{100 \times 500 \times 10^6} = 20\Omega$$

$$\frac{1}{\omega C_2} = \frac{1}{100 \times 2500 \times 10^6} = 4\Omega$$

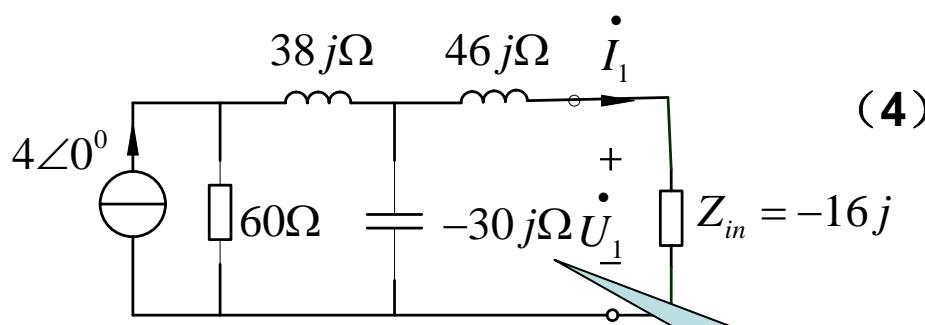


(2)

$$Z_{in} = 2^2 \times (-4j) = -16j\Omega$$

$$46j - 16j = 30j\Omega$$

$$\dot{U}_1 = 4\angle 0^0 \times 60 = 240\angle 0^0$$



(4)

$$\dot{I}_L = \frac{\dot{U}}{30j} = \frac{240\angle 0^0}{30\angle 90^0} = 8\angle -90^0$$

$$\dot{U}_1 = Z_{in} \dot{I}_L = 8\angle -90^0 \times 16\angle -90^0 = 128\angle -180^0$$

(5)

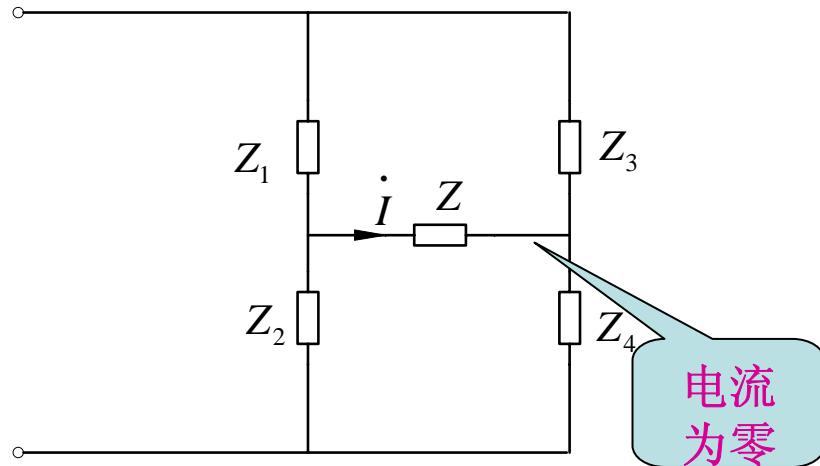
$$\dot{U}_c = -0.5\dot{U}_1 = 64\angle 0^0$$

$$\dot{I}_C = -2\dot{I}_L = 16\angle 90^0$$

伏安  
关系

并联谐振  
断路

## 交流电桥平衡



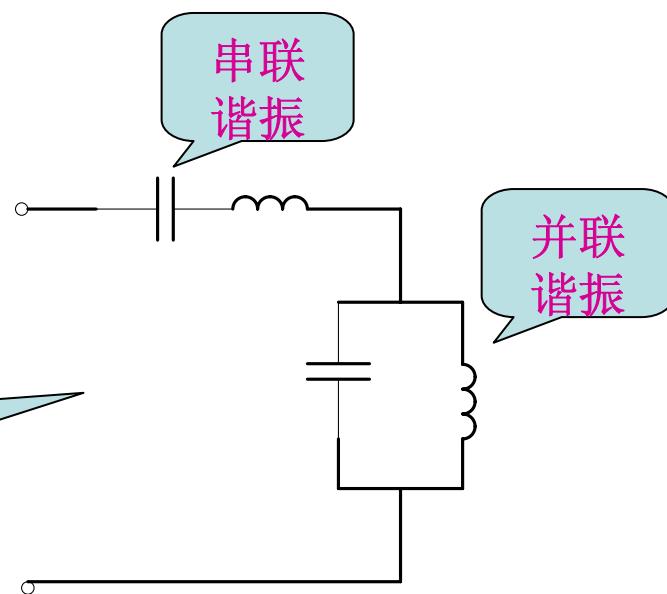
$$Z_1 \times Z_4 = Z_2 \times Z_3$$

多个谐振频率

串联  
谐振

并联  
谐振

串联  
谐振

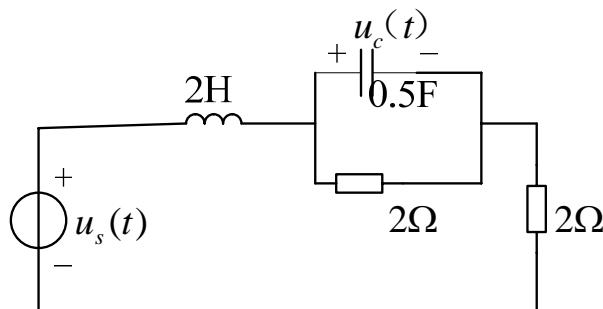


# 第六章 线性动态电路的时域分析

## 一、输入输出方程

根据两类约束  $i_c = C \frac{du_c}{dt}$   $u_l = L \frac{di_l}{dt}$

6-1



$$i_1 = \frac{u_c(t)}{2} \quad i_c = C \frac{du_c(t)}{dt}$$

$$i_l = i_1 + i_c = \frac{u_c(t)}{2} + C \frac{du_c(t)}{dt}$$

$$u_s = L \frac{di_l(t)}{dt} + u_c(t) + 2 \times i_l$$

$$u_s = L \frac{d(\frac{u_c(t)}{2} + C \frac{du_c(t)}{dt})}{dt} + u_c(t) + 2 \times \frac{u_c(t)}{2} + 2 \times C \frac{du_c(t)}{dt}$$

$$u_s = \frac{du_c(t)}{dt} + \frac{d^2 u_c(t)}{dt^2} + 2u_c(t) + \frac{du_c(t)}{dt}$$

$$= 2 \frac{du_c(t)}{dt} + \frac{d^2 u_c(t)}{dt^2} + 2u_c(t)$$

## 二、三要素法

- 1) 根据换路前稳态电路求  $u_c(0_-)$   $i_L(0_-)$  根据换路定则确定
- 2) 根据零正时刻电路求  $y(0_+)$   $u_c(0_+) = u_c(0_-)$
- 3) 根据换路后稳态电路求  $y(\infty)$   $i_L(0_+) = i_L(0_-)$
- 4) 将换路后的电路化为无源电路，求解储能元件两端的等效电阻  $R_{eq}$  并求时间常数。
- 5) 代入三要素公式

$$y(t) = y(\infty) + [y(0_+) - y(\infty)] e^{-\frac{t}{\tau}}$$

待求电压或电流的初始值

待求电压或电流

待求电压或电流的稳态值

电路时间常数

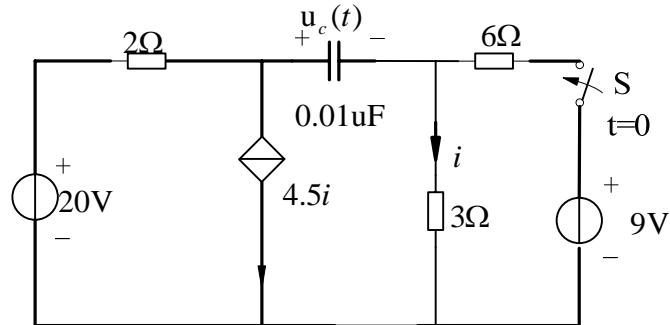
电阻电容电路

$$\tau = R_{eq} \times C$$

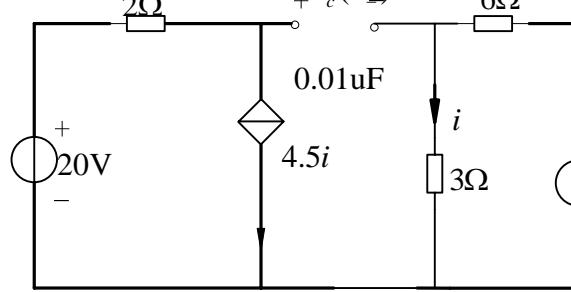
电阻电感电路

$$\tau = \frac{l}{R_{eq}}$$

例：求开关闭合后电容电压和电流



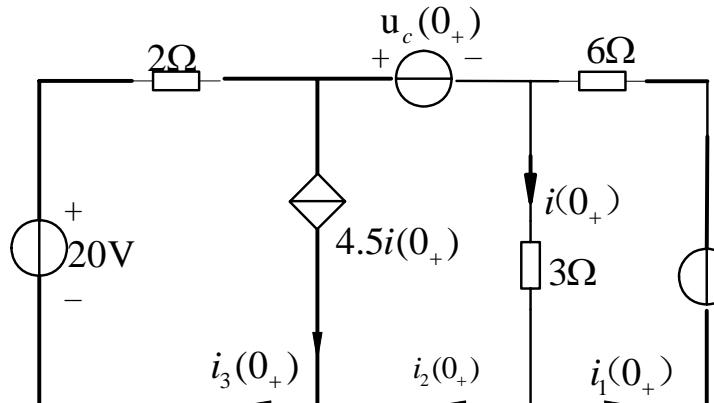
解1) 求电容电压  $u_c(0_-)$



换路前直流稳态电路

$$u_c(0_-) = 20V$$

2) 求  $u_c(0_+)$   $i(0_+)$



$$u_c(0_+) = u_c(0_-) = 20V$$

零正时刻电路

$$i_1(0_+) = (9 - 3i(0_+))/6$$

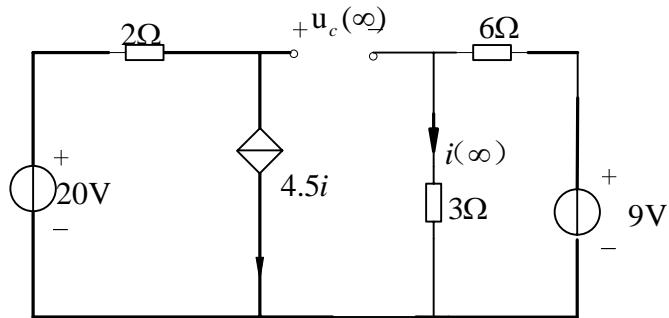
$$i_2(0_+) = i(0_+) - i_1(0_+) = 1.5i(0_+) - 1.5$$

$$i_3(0_+) = 4.5i(0_+) + i_2(0_+) = 6i(0_+) - 1.5$$

$$-20 + 2i_3(0_+) + 20 + 3i(0_+) = 0$$

$$12i(0_+) + 3i(0_+) = 3 \quad i(0_+) = 0.2A$$

3) 求  $u_c(\infty)$   $i(\infty)$



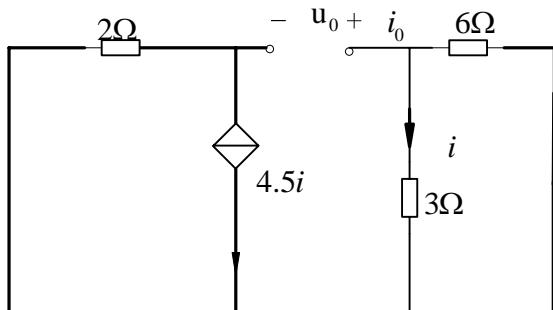
$$u_c(\infty) + i(\infty) \times 3 - 20 + 4.5i(\infty) \times 2 = 0$$

$$u_c(\infty) = 8V$$

$$i(\infty) = \frac{9}{6+3} = 1A$$

换路后直流稳态电路

4) 求输入电阻



$$i_0 = i + \frac{3i}{6} = 1.5i \quad u_0 = 3i + (i_0 + 4.5i) \times 2 = 15i$$

$$R = \frac{u_0}{i_0} = 10\Omega \quad \tau = R \times C = 10 \times 0.01 = 0.1S$$

换路后无源电路

5) 代入三要素公式求响应

$$u_c(t) = u_c(\infty) + [u_c(0_+) - u_c(\infty)] e^{\frac{-t}{\tau}} = 8 + 12e^{-10t}V$$

$$i(t) = i(\infty) + [i(0_+) - i(\infty)] e^{\frac{-t}{\tau}} = 1 - 0.8e^{-10t}A$$

### 三、一阶动态电路的线性特性

全响应=零输入响应+零状态相应

零输入响应

当  $u_c(0_-) i_L(0_-)$  变化k倍时，零输入响应也增加k倍

零状态响应

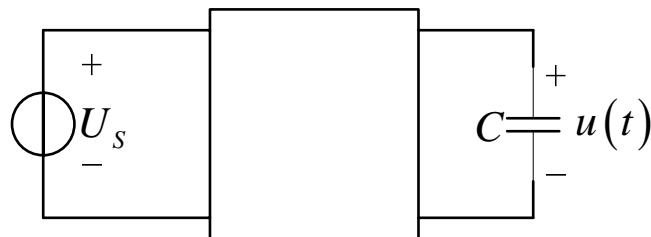
当输入激励变化k倍，零状态响应也增加k倍

注意齐次性

6-17

已知  $u_s(t) = 10\delta(t)$  时  $u_c = (3 + 5e^{-3t})\delta(t)$

求  $u_s(t) = 5\delta(t)$  时的全响应



解  $u_s(t) = 10\delta(t)$   $u_c = (3 + 5e^{-3t})\delta(t)$

$$u_c(\infty) = (3 + 5e^{-3 \times \infty}) = 3$$

$$u_c(0_+) = (3 + 5e^{-3 \times 0_+}) = 8$$

根据零状态齐次性  $u_s(t) = 5\delta(t)$   $u_c(\infty) = 1.5$

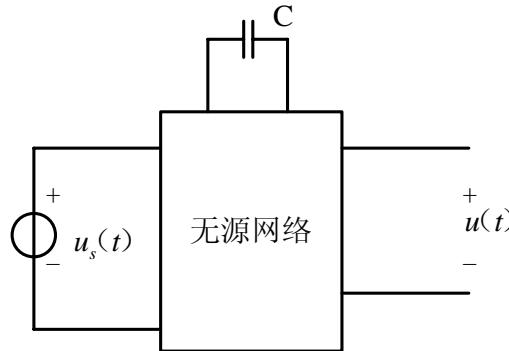
$$u_c(t) = (1.5 - 1.5e^{-3t}) + 8e^{-3t} = 1.5 + 6.5e^{-3t}$$

零输入

零状态

电容电压电感电流的全响应方程可以分解出零输入和零状态 响应，其它不可以。

6-17



解  $u_s(t) = 10\varepsilon(t)$

$$10 + 4e^{-t} = u_{\text{cx}} + u_{\text{cf}}$$

根据零状态齐次性

$$u_s(t) = 5\varepsilon(t)$$

$$5 + 6e^{-t} = u_{\text{cx}} + 10/5 \times u_{\text{cf}}$$

当  $u_s(t) = 10\varepsilon(t)$  作用时，响应  $u(t) = 10 + 4e^{-t}\varepsilon(t)$

当  $u_s(t) = 5\varepsilon(t)$  作用时，响应  $u(t) = 5 + 6e^{-t}\varepsilon(t)$

则

求零输入响应

$$u(t) = u_{\text{cx}} + u_{\text{cf}} \xrightarrow{\text{零输入}} \text{零状态}$$

$$u_{\text{cx}} = 8e^{-t}\varepsilon(t)$$

当  $u_s(t) = 10\varepsilon(t)$   $u_c(0_-) = 20V$

$$u(t) = 10 + 4e^{-t}\varepsilon(t)$$

当  $u_c(0_-) = 30V$   $u(t) = 10 + 8e^{-t}\varepsilon(t)$

求零状态响应

$$u(t) = u_{\text{cx}} + u_{\text{cf}} \xrightarrow{\text{零输入}} \text{零状态}$$

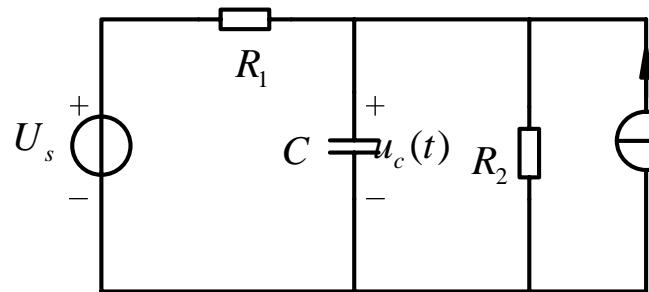
解 当  $u_c(0_-) = 20V$   $10 + 4e^{-t} = u_{\text{cx}} + u_{\text{cf}}$

根据零输入齐次性

当  $u_c(0_-) = 30V$

$$10 + 8e^{-t} = 30/20 \times u_{\text{cx}} + u_{\text{cf}}$$

$$u_{\text{cf}} = 10 - 4e^{-t}$$



电路初始状态保持不变

$$I_s U_s = 1 \quad I_s = 0 \quad u_c(t) = 0.5 + 2e^{-2t}$$

$$U_s = 0 \quad I_s = 1 \quad u_c(t) = 2 + 0.5e^{-2t}$$

求电阻和电容值。

求当  $U_s = 1 \quad I_s = 1$  电路的响应。

---

1) 当  $U_s = 1 \quad I_s = 0 \quad u_c(t) = 0.5 + 2e^{-2t}$

则  $u_c(\infty) = 0.5 \quad u_c(0_+) - u_c(\infty) = 2 \quad u_c(0_+) = 2.5 \quad U_c(\infty) = \frac{R_2}{R_1 + R_2} U_s$

---

2) 当  $U_s = 0 \quad I_s = 1 \quad u_c(t) = 2 + 0.5e^{-2t}$

则  $u_c(\infty) = 2 \quad u_c(0_+) - u_c(\infty) = 0.5 \quad u_c(0_+) = 2.5 \quad U_c(\infty) = \frac{R_2 R_1}{R_1 + R_2} I_s$

$$\tau = R_1 // R_2 \times C = 0.5$$

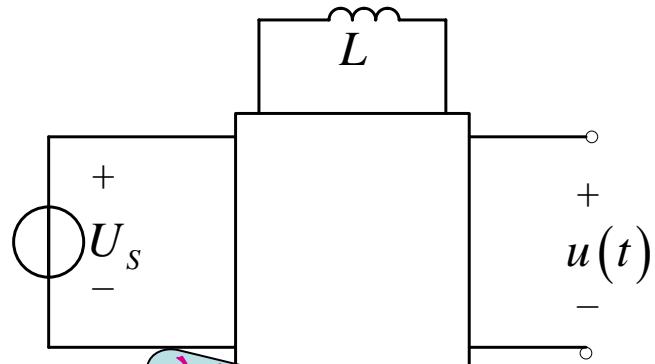

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3)  $u_c'(\infty) = 0.5V \quad u_c''(\infty) = 2V \quad u_c(\infty) = u_c'(\infty) + u_c''(\infty) = 2.5V$

$$u_c(t) = u_c(\infty) + [u_c(0_+) - u_c(\infty)] e^{-2t} = 2.5V$$

已知  $u_s = 8\varepsilon(t)$  时  $u = (3 + 2e^{-3t})\varepsilon(t)$   $u_s = 16\varepsilon(t)$  时  $u = (6 + 5e^{-3t})\varepsilon(t)$

求如图所示激励的全响应

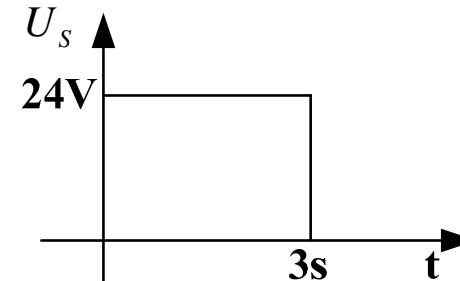


解  
题

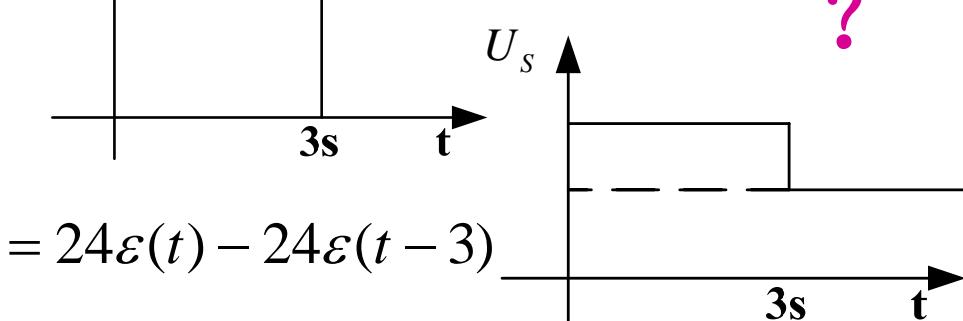
零状态

$$3 + 2e^{-3t} = u_x + u_h \quad \rightarrow$$

$$u_s = 24\varepsilon(t) - 24\varepsilon(t-3)$$



?



$$u_x = 2(3 + 2e^{-3t}) - (6 + 5e^{-3t}) = -e^{-3t}$$

(1)

$$6 + 5e^{-3t} = u_x + 2u_h \quad u_h = 3 + 3e^{-3t}$$

$$(2) u_h = 3u_h = (9 + 9e^{-3t})\varepsilon(t) \quad \text{零状态  
线性}$$

$$u_h = (9 + 9e^{-3t})\varepsilon(t) - (9 + 9e^{-3(t-3)})\varepsilon(t-3)$$

$$u = u_h + u_x = (9 + 9e^{-3t})\varepsilon(t) - (9 + 9e^{-3(t-3)})\varepsilon(t-3) - e^{-3t}$$

## 单位冲激响应 $h(t)$

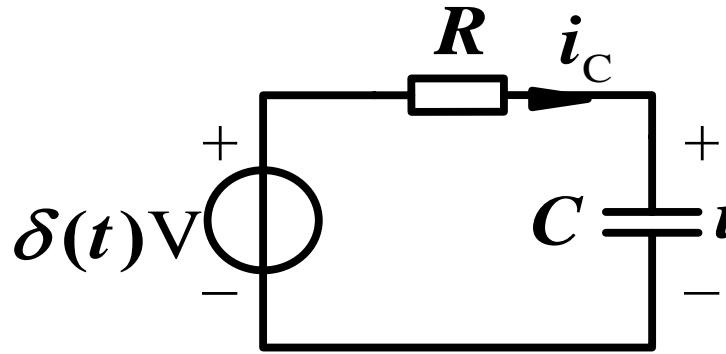
(1) 定义：电路对单位冲激输入的零状态响应称为单位冲激响应，并用 $h(t)$ 表示。

(2) 求解方法

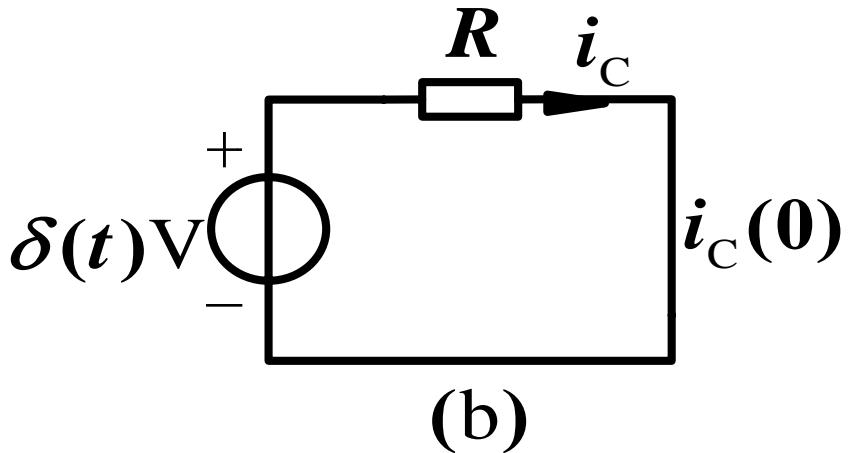
① 求在 ( $0_- \sim 0_+$ ) 瞬间，冲激激励作用下，在零状态电路中建立的初始状态（因电容电流和电感电压跃变，不满足换路定则）。

② 求 ( $t \geq 0_+$ ) 时，初始状态作用下的零输入响应。

【例1】求冲激响应 $u_C(t)$ 和 $i_C(t)$ 。



(a)



(b)

解：分两个时间段分析求冲激响应。

(1) 在 $0_- \sim 0_+$ 间，电容充电，电容等效为短路。

$$i_C(0) = \frac{\delta(t)}{R}$$

根据电容电压电流关系

$$u_c(0_+) = u_c(0_-) + \frac{1}{C} \int_{0_-}^{0_+} i_c(t) dt = \frac{1}{C} \int_{0_-}^{0_+} \frac{\delta(t)}{R} dt = \frac{1}{RC}$$

(2) 在  $t \geq 0_+$  间，电容对电阻放电，电路为零输入响应。

$$\therefore h_u(t) = u_C(t) = u_C(0_+) e^{-\frac{t}{RC}} \varepsilon(t) = \frac{1}{RC} e^{-\frac{t}{RC}} \varepsilon(t) (\text{V})$$

$$\therefore h_u(t) = u_C(t) = u_C(0_+) e^{-\frac{t}{RC}} \varepsilon(t) = \frac{1}{RC} e^{-\frac{t}{RC}} \varepsilon(t)$$

$$\therefore i_c(t) = C \frac{du_C(t)}{dt}$$

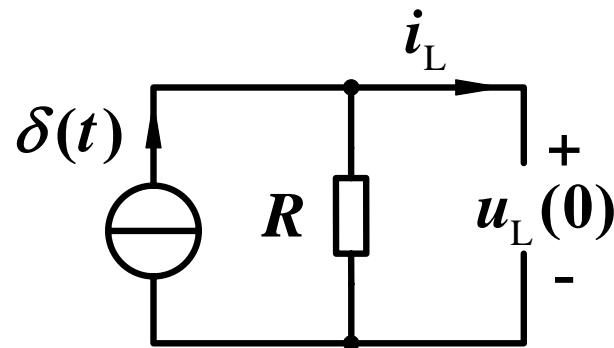
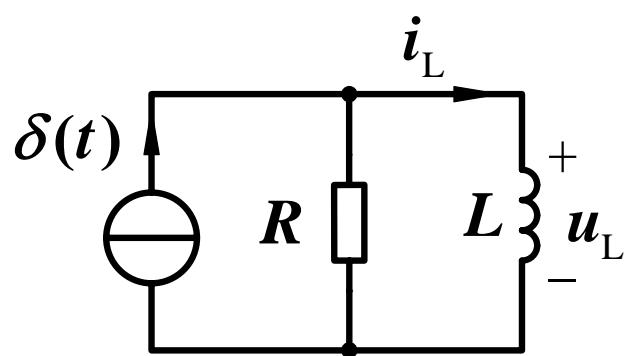
作为函数求导

$$= C \frac{1}{RC} \left[ \left( -\frac{1}{RC} \right) e^{-\frac{t}{RC}} \varepsilon(t) + e^{-\frac{t}{RC}} \frac{d\varepsilon(t)}{dt} \right]$$

$$= -\frac{1}{R^2 C} e^{-\frac{t}{RC}} \varepsilon(t) + \frac{1}{R} \left[ e^{-\frac{t}{RC}} \delta(t) \right]_{t=0}$$

$$= -\frac{1}{R^2 C} e^{-\frac{t}{RC}} \varepsilon(t) + \frac{1}{R} \delta(t)$$

【例2】求冲激响应  $i_L(t)$  和  $u_L(t)$ 。



解：分两个时间段分析，求冲激响应。

(1) 在  $0_- \sim 0_+$  间，电感等效为开路。

$$u_L(0) = \delta(t)R$$

(2) 根据电感电压电流关系。

$$i_L(0_+) = i_L(0_-) + \frac{1}{L} \int_{0_-}^{0_+} u_L dt = \frac{1}{L} \int_{0_-}^{0_+} R\delta(t) dt = \frac{R}{L}$$

$$\therefore i_L(0_+) = \frac{R}{L}$$

(2) 在  $t \geq 0_+$  间，电容对电阻放电，电路为零输入响应。

$$\therefore h_i(t) = i_L(t) = i_L(0_+) e^{-\frac{t}{RC}} \varepsilon(t) = \frac{R}{L} e^{-\frac{Rt}{L}} \varepsilon(t)$$

$$\therefore h_u(t) = u_L(t) = L \frac{di_L(t)}{dt}$$

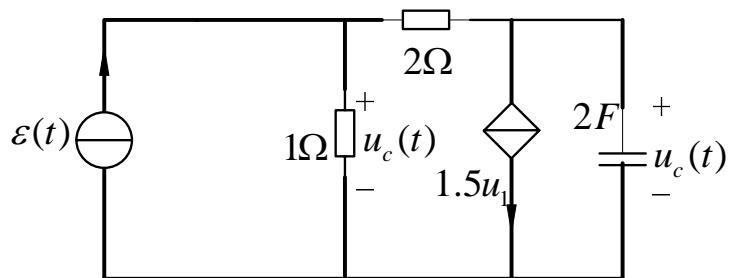
作为函数求导

$$= L \frac{R}{L} \left[ \left( -\frac{R}{L} \right) e^{-\frac{Rt}{L}} \varepsilon(t) + e^{-\frac{Rt}{L}} \frac{d\varepsilon(t)}{dt} \right]$$

$$= -\frac{R^2}{L} e^{-\frac{Rt}{L}} \varepsilon(t) + R \left[ e^{-\frac{Rt}{L}} \delta(t) \right]_{t=0}$$

$$= -\frac{R^2}{L} e^{-\frac{Rt}{L}} \varepsilon(t) + R \delta(t)$$

## 6-19 求电路单位阶跃响应



1) 求初始电容电压

$$u_c(0_-) = QV \quad u_c(0_-) = u_c(0_+) = 0V$$

单位  
阶跃

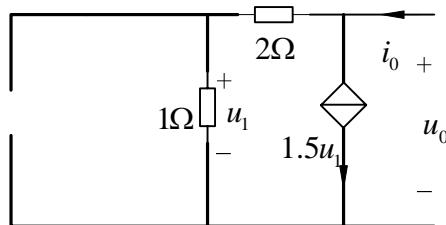
2) 求稳态值

$$\varepsilon(t) = \frac{u_1}{1} + 1.5u_1 = 1 \quad u_1 = \frac{1}{2.5}V$$

$$u_c(\infty) - u_1 + 2 \times 1.5u_1 = 0V$$

$$u_c(\infty) = -2u_1 = -0.8V$$

3) 求时间常数



$$u_0 = 2u_1 + u_1 = 3u_1 \quad i_0 = 1.5u_1 + u_1/1 = 2.5u_1$$

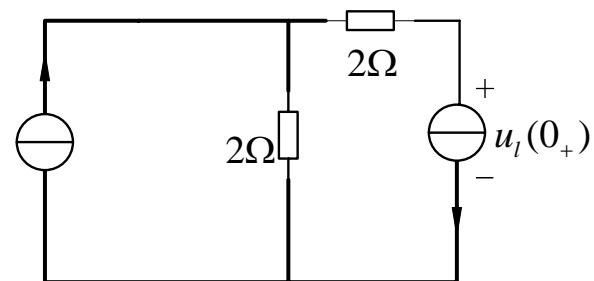
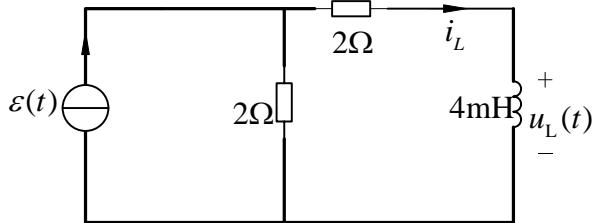
$$R = \frac{3u_1}{2.5u_1} = \frac{6}{5}\Omega \quad \tau = R \times C = 1.2 \times 2 = 2.4S$$

4)

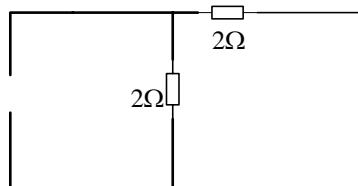
$$u(t) = u(\infty) + [u(0_+) - u(\infty)]e^{\frac{-t}{\tau}}$$

$$= \left( -0.8 + 0.8e^{-\frac{t}{2.4}} \right) \varepsilon(t)V$$

求电路单位阶跃响应



3) 求时间常数



$$R = 4\Omega$$

$$\tau = L/R = 0.004 / 4 = 0.001S$$

1) 求初始电流

$$i_L(0_-) = 0$$

$$i_L(0_+) = i_L(0_-) = 0$$

2) 求初始值

$$u_l(0_+) = 1 \times 2 = 2V$$

3) 求稳态值  $u_l(\infty) = 0V$

$$4) \quad u(t) = 2e^{-1000t} \varepsilon(t)V$$

单位  
阶跃

单位  
冲击  
为单  
位阶  
跃的  
导数

$$u_\delta(t) = \frac{du_\varepsilon}{dt} = \frac{d2e^{-1000t}\varepsilon(t)}{dt}$$
$$= 2\delta(t) - 2000e^{-1000t}\varepsilon(t)$$

## 6-20 求图示电路电感电压单位冲击响应

1) 根据图1求初始电压

$$u_L(0) = 2\delta(t)$$

$$i_L(0_+) = \frac{1}{L} \int_{0^-}^{0^+} u_L(0) dt = \frac{1}{L} \int_{0^-}^{0^+} 2\delta(t) dt = 500A$$

2) 由图2求初始值

$$\begin{aligned} u_l(0_+) &= -500 \times (2 + 2) \\ &= -2000V \end{aligned}$$

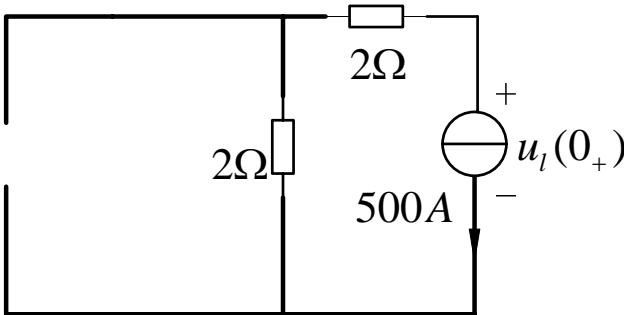


图2

3) 求稳态值

$$u_l(\infty) = 0V$$

$$4) \quad u(t) = 2\delta(t) - 2000e^{-1000t}\varepsilon(t)V$$

补充电感电压在零时刻值

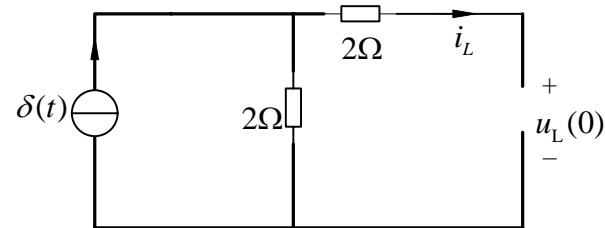
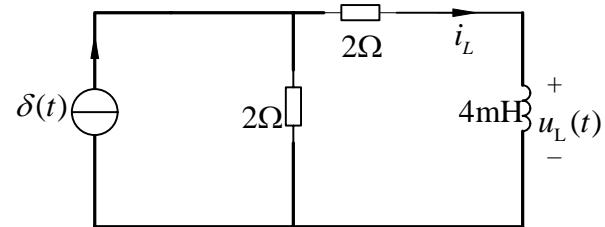


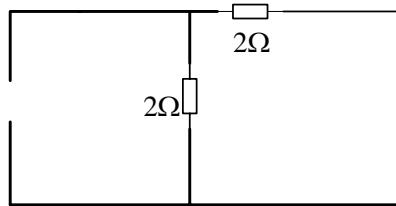
图1

电感开路，电容短路

4) 求时间常数

$$R = 4\Omega$$

$$\tau = L/R = 0.004 / 4 = 0.001S$$



# 假二阶电路

电路等效为两个一阶动态电路，根据三要素法分别求解。

## 四、二阶动态电路的方程及三种响应

列写电路动态方程，根据特征根的取值来判断其相应特性。

标准形式

$$\frac{d^2y}{dt^2} + 2\alpha \frac{dy}{dt} + \omega^2 y = 0$$

$$\lambda^2 + 2\alpha\lambda + \omega^2 = 0$$

特征方程

$$\lambda_1 = -\alpha + \sqrt{\alpha^2 - \omega^2} \quad \lambda_2 = -\alpha - \sqrt{\alpha^2 - \omega^2}$$

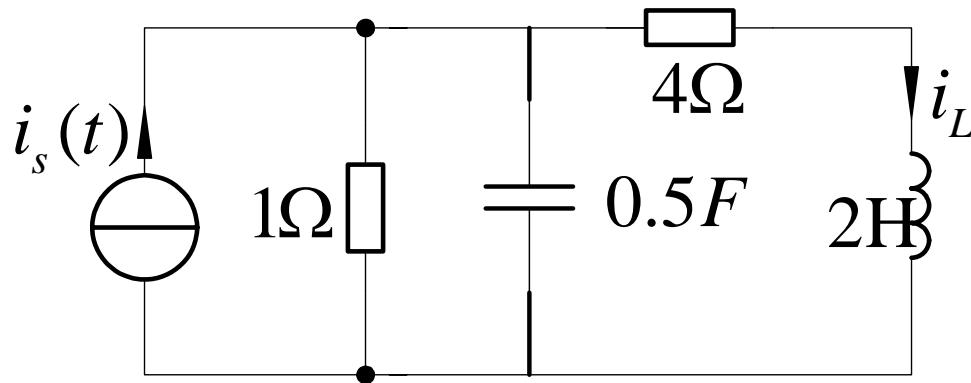
两个不相等的实数根，过阻尼；  $\alpha > \omega$

两个相等的实数根，临界阻尼；  $\alpha = \omega$

两个不相等的复数根，欠阻尼；  $\alpha < \omega$

两个共轭纯虚数根，无阻尼；  $\alpha = 0$

列写以电感电流为输出变量的输入输出方程。



$$\frac{di_L^2}{dt^2} + 4 \frac{di_L}{dt} + 5i_L = i_s \quad \lambda^2 + 4\lambda + 5 = 0$$

$$a\lambda^2 + b\lambda + c = 0$$

$$\lambda_{12} = \frac{-b \pm \sqrt{b^2 - 4ac}}{2a}$$

激励  
为零

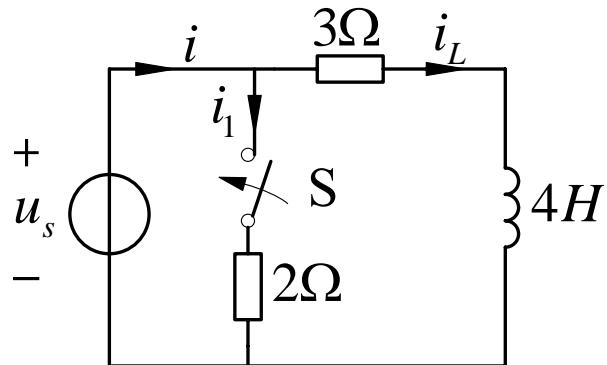
$$\lambda^2 + 2\alpha\lambda + \omega^2 = 0 \quad \alpha = 2 \quad \omega = \sqrt{5}$$

$\alpha < \omega$  两个不相等的复数根，欠阻尼；

# 第六章 线性动态电路的时域分析

## 非直流量激励下的暂态响应分析

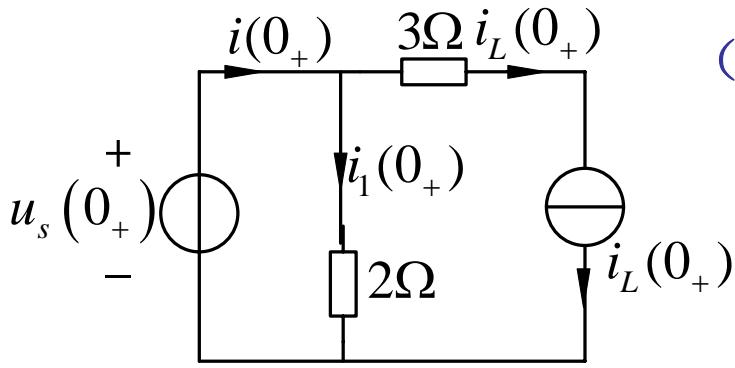
求换路后的各电流初始值  $u_s = 10 \cos(t)V$



(1) 求换路前稳态值

Circuit diagram for finding the initial value of current  $i_L$  at  $t=0$ . It shows a voltage source  $\dot{U}_s$ , a  $3\Omega$  resistor, and a  $2\Omega$  resistor in series.

$$\dot{i}_L = \dot{U}_s / Z = \frac{\dot{U}_{sm}}{Z} = \frac{10 \angle 0^\circ}{5 \angle 53^\circ} = 2 \angle -53^\circ$$
$$\dot{i}_L = 2 \cos(t - 53^\circ)$$
$$i_L(0_+) = i_L(0_-) = 2 \cos(0_- - 53^\circ) = 1.2$$



(2) 求初始值

$$u_s(0_+) = 10 \cos(0_+) = 10 \quad i_1(0_+) = \frac{u_s(0_+)}{2} = \frac{10}{2} = 5$$

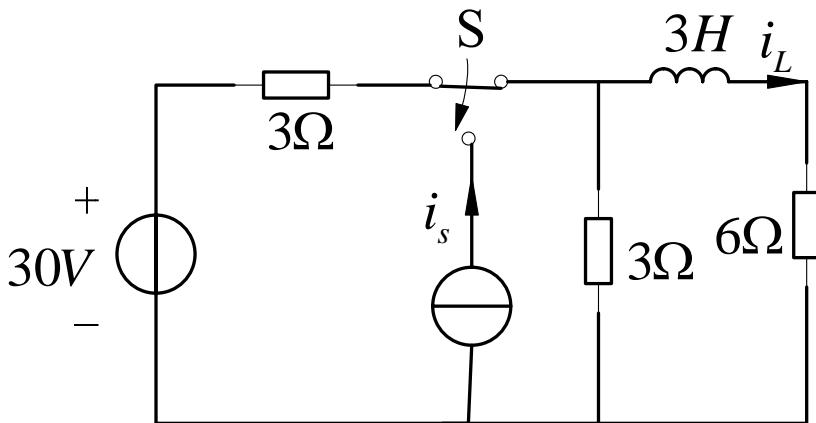
$$i(0_+) = i_L(0_+) + i_1(0_+) = 5 + 1.2 = 6.2A$$

求换路后的电感电流  $i_s = 9\sqrt{2} \cos 3t$

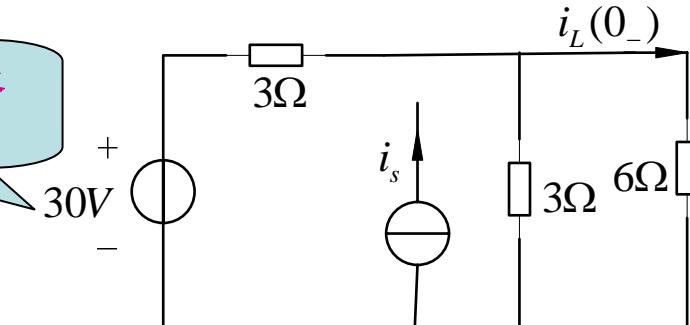
### (1) 求换路前稳态值

$$i_L(0_-) = \frac{30}{3+6//3} \times \frac{3}{3+6} = 2A$$

$$i_L(0_+) = i_L(0_-) = 2A$$



换路前稳态电路

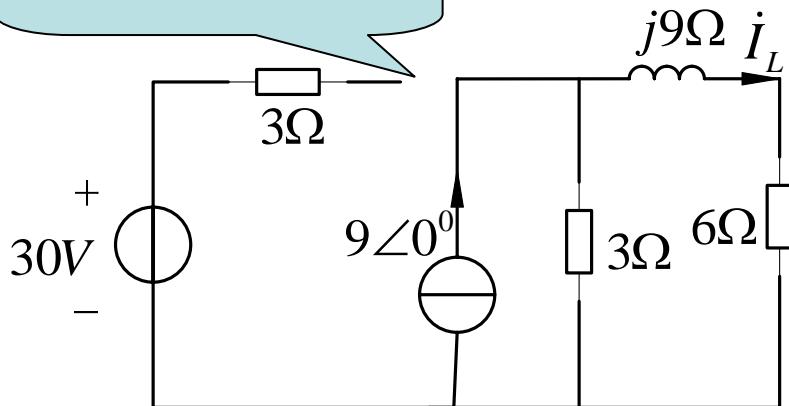


### (2) 求换路后稳态值（相量分析法）

$$\dot{I}_L = \frac{3}{6+9j+3} 9\angle 0^\circ = 1.5\sqrt{2}\angle -45^\circ$$

$$i_L = 3\cos(3t - 45^\circ)$$

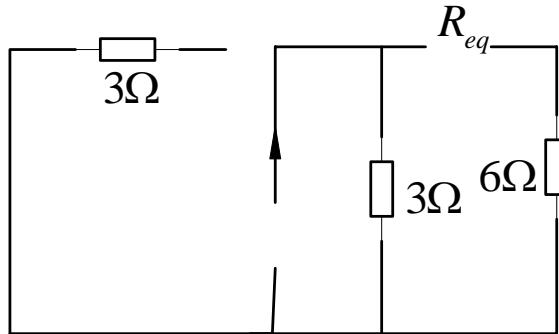
换路后稳态电路



### (3) 求时间常数

$$R_{eq} = 9\Omega$$

$$\tau = \frac{L}{R_{eq}} = \frac{1}{3}s$$



### (4) 代入三要素公式

稳态值在零正时  
刻的值

$$i_L(t) = i_L(\infty) + [i_L(0_+) - i_L(\infty)|_{(t=0_+)}] e^{-\frac{t}{\tau}}$$

$$i_L(t) = 3\cos(3t - 45^\circ) + [2 - 3\cos(3 \times 0_+ - 45^\circ)] e^{-3t}$$

$$= 3\cos(3t - 45^\circ) + [2 - 3\cos(3 \times 0_+ - 45^\circ)] e^{-3t}$$

$$= 3\cos(3t - 45^\circ) - 0.121e^{-3t}$$

# 答疑时间及地点

地点： 教11A420

**11月27 日全天**

**11月28 日下午全天**

**考试时间11月29日（周六）**